#### Capacity Achieving Codes: There and Back Again

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#### Outline

Introduction

Factor Graphs

Message Passing

Applications of Factor Graphs

Applications of EXIT Curves

Spatially-Coupled Factor Graphs

Universality for Multiuser Scenarios

Abstract Formulation of Threshold Saturation

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- Applications of EXIT Curves
- Spatially-Coupled Factor Graphs
- Universality for Multiuser Scenarios
- Abstract Formulation of Threshold Saturation

#### Capacity of Point-to-Point Communication

$$X \longrightarrow P_{Y|X} \longrightarrow Y$$

Coding for Discrete-Time Memoryless Channels

- Transition probability:  $P_{Y|X}(y|x)$  for  $x \in \mathcal{X}$  and  $y \in \mathcal{Y}$
- Transmit a length-n codeword  $\underline{x} \in \mathcal{C} \subset \mathcal{X}^n$
- $\blacktriangleright$  Decode to most likely codeword given received y

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- Channel Capacity introduced by Shannon in 1948
  - ▶ Random code of rate  $R \triangleq \frac{1}{n} \log_2 |\mathcal{C}|$  (bits per channel use)
  - As  $n \to \infty$ , reliable transmission possible if R < C with

$$C \triangleq \max_{p(x)} I(X;Y)$$

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$$H\underline{x} = \begin{bmatrix} H_{\mathcal{E}} & H_{\mathcal{E}^c} \end{bmatrix} \begin{bmatrix} \underline{x}_{\mathcal{E}} \\ \underline{y}_{\mathcal{E}^c} \end{bmatrix} = \underline{0} \quad \Leftrightarrow \quad H_{\mathcal{E}}\underline{x}_{\mathcal{E}} = -H_{\mathcal{E}^c}\underline{y}_{\mathcal{E}^c}$$

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- ▶ One can achieve capacity by drawing *H* uniformly at random!

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- 1960: Reed-Solomon codes

#### Achieving Capacity in Practice

But, more than 35 years passed before we could:

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- ▶ 1999-2011: Understanding LDPC convolutional codes and coupling

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- EXtrinsic Information Transfer (EXIT) Curves

# Applications of These Tools

- Error-Correcting Codes
  - Random code defined by random factor graph
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  - Non-rigorous analysis via the cavity method
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- Compressed Sensing
  - Random measurement matrix defined by random factor graph
  - Low-complexity reconstruction via message passing
  - Schemes provably achieve the information-theoretic limit!

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- ► The solution of the simpler problem often provides insight that allows one to crack the harder problem.
- To achieve channel capacity in practice, we now know that a good "easy" problem would have been:
  - "Design a code that achieves capacity on the BEC and is encodable and decodable in quasi-linear time"

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# Factor Graphs

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  - Bipartite graph with variables  $x_1, \ldots, x_n$  and factors  $f_1, \ldots, f_m$
- Consider random variables  $(X_1, X_2, \ldots, X_4) \in \mathcal{X}^4$  and Y where:

$$P(x_1, x_2, x_3, x_4) \triangleq \mathbb{P}(X_1 = x_1, X_2 = x_2, \dots, X_4 = x_4 | Y = y)$$
  
\$\approx f(x\_1, x\_2, x\_3, x\_4)\$  
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• Given Y = y, this describes a Markov chain whose factor graph is



 $\blacktriangleright$  Let  $A,B,S\subset [n]$  be disjoint subsets of VNs in factor graph G

If S separates A from B (i.e., there is no path in G from A to B that avoids S), then we have X<sub>A</sub> ⊥⊥ X<sub>B</sub> | X<sub>S</sub>

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Sketch of Proof:

- ▶ Fixing  $X_S = x_S$  separates the FG into disjoint components
- Groups of VNs in different components are independent
- $X_A \perp \!\!\!\perp X_B$  because A and B are in different components

### Inference via Marginalization

• Marginalizing out all variables except  $X_1$  gives

$$\mathbb{P}(X_1 = x_1 | Y = y) \propto g_1(x_1) \triangleq \sum_{(x_2, \dots, x_4) \in \mathcal{X}^3} f(x_1, x_2, x_3, x_4)$$

▶ Thus, the maximum a posteriori decision for  $X_1$  given Y = y is

$$\hat{x}_1 = \arg \max_{x_1 \in \mathcal{X}} \sum_{(x_2, \dots, x_4) \in \mathcal{X}^3} f(x_1, x_2, x_3, x_4)$$

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Marginalization is efficient for tree-structured factor graphs

For the Markov chain, roughly  $5 |\mathcal{X}|^2$  operations required

$$g_1(x_1) = \sum_{x_2 \in \mathcal{X}} f_1(x_1, x_2) \sum_{x_3 \in \mathcal{X}} f_2(x_2, x_3) \sum_{x_4 \in \mathcal{X}} f_3(x_3, x_4)$$

• Consider a random vector  $(X_1, X_2, \ldots, X_6) \in \mathcal{X}^6$  where

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• Brute force marginal requires  $|\mathcal{X}|^5$  operations for each  $x_1 \in \mathcal{X}$ :

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• If f factors as follows, then the marginalization can be simplified:

 $f(x_1, x_2, x_3, x_4, x_5, x_6) = f_1(x_1, x_2, x_3) f_2(x_1, x_4, x_6) f_3(x_4) f_4(x_4, x_5)$ 

$$=\sum_{x_2^6} f_1(x_1, x_2, x_3) f_2(x_1, x_4, x_6) f_3(x_4) f_4(x_4, x_5)$$

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$$=\sum_{x_2^5} f_1(x_1, x_2, x_3) f_3(x_4) f_4(x_4, x_5) \left[\sum_{x_6} f_2(x_1, x_4, x_6)\right]$$

$$= \sum_{x_2^6} f_1(x_1, x_2, x_3) f_2(x_1, x_4, x_6) f_3(x_4) f_4(x_4, x_5)$$
  
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$$= \sum_{x_2^4} f_1(x_1, x_2, x_3) f_3(x_4) \left[ \sum_{x_5} f_4(x_4, x_5) \right] \left[ \sum_{x_6} f_2(x_1, x_4, x_6) \right]$$

$$\begin{split} &= \sum_{x_2^6} f_1(x_1, x_2, x_3) f_2(x_1, x_4, x_6) f_3(x_4) f_4(x_4, x_5) \\ &= \sum_{x_2^5} f_1(x_1, x_2, x_3) f_3(x_4) f_4(x_4, x_5) \left[ \sum_{x_6} f_2(x_1, x_4, x_6) \right] \\ &= \sum_{x_2^4} f_1(x_1, x_2, x_3) f_3(x_4) \left[ \sum_{x_5} f_4(x_4, x_5) \right] \left[ \sum_{x_6} f_2(x_1, x_4, x_6) \right] \\ &= \sum_{x_2^3} f_1(x_1, x_2, x_3) \left[ \sum_{x_4} f_3(x_4) \left[ \sum_{x_5} f_4(x_4, x_5) \right] \left[ \sum_{x_6} f_2(x_1, x_4, x_6) \right] \right] \end{split}$$

For example, we can write  $g_1(x_1)$  as:

$$= \sum_{x_2^6} f_1(x_1, x_2, x_3) f_2(x_1, x_4, x_6) f_3(x_4) f_4(x_4, x_5)$$

$$= \sum_{x_2^5} f_1(x_1, x_2, x_3) f_3(x_4) f_4(x_4, x_5) \left[ \sum_{x_6} f_2(x_1, x_4, x_6) \right]$$

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$$= \sum_{x_2^3} f_1(x_1, x_2, x_3) \left[ \sum_{x_4} f_3(x_4) \left[ \sum_{x_5} f_4(x_4, x_5) \right] \left[ \sum_{x_6} f_2(x_1, x_4, x_6) \right] \right]$$

$$= \sum_{x_2} \left[ \sum_{x_3} f_1(x_1, x_2, x_3) \right] \left[ \sum_{x_4} f_3(x_4) \left[ \sum_{x_5} f_4(x_4, x_5) \right] \left[ \sum_{x_6} f_2(x_1, x_4, x_6) \right] \right]$$

This implementation requires roughly  $2\left|\mathcal{X}\right|^3+5\left|\mathcal{X}\right|^2$  operations







$$g_1(x_1) = \sum_{x_2^3} f_1(x_1, x_2, x_3) \left[ \sum_{x_4} f_3(x_4) f'_4(x_4) f'_2(x_1, x_4) \right]$$



$$g_1(x_1) = \sum_{x_2} \left[ \sum_{x_3} f_1(x_1, x_2, x_3) \right] f_2''(x_1)$$



$$g_1(x_1) = \left[\sum_{x_2} f_1'(x_1, x_2)\right] f_2''(x_1)$$



$$g_1(x_1) = f_1''(x_1)f_2''(x_1)$$

### Constraint Satisfaction and Zero-One Factors

• A non-negative function  $f: \mathcal{X}^n \to \mathbb{R}$  defines a distribution on  $\mathcal{X}^n$ :

$$P(\underline{x}) \triangleq \mathbb{P}(X_1 = x_1, \dots, X_n = x_n)$$
$$= \frac{1}{Z} f(\underline{x}) \triangleq \frac{1}{Z} \prod_{a=1}^m f_a(\underline{x}_{\partial a}),$$

▶ where  $\underline{x}_{\partial a}$  is the subvector of variables involved in factor a▶ and  $Z \triangleq \sum_{\underline{x}} f(\underline{x})$  is called the partition function

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- For Constraint Satisfaction Problems (CSPs)
  - All factors  $f_a(\underline{x}_{\partial a})$  take values in  $\{0, 1\}$
  - The set of valid configurations is  $\{\underline{x} \in \mathcal{X}^n | f(\underline{x}) = 1\}$
  - $\blacktriangleright$  Thus, Z equals the number of valid configurations
  - $P(\underline{x})$  is uniform over the set of valid configurations

## Outline

Introduction

Factor Graphs

Message Passing

Applications of Factor Graphs

Applications of EXIT Curves

Spatially-Coupled Factor Graphs

Universality for Multiuser Scenarios

Abstract Formulation of Threshold Saturation

- Factor Graph  $G = (V \cup F, E)$ 
  - Variable nodes V, Factor nodes F
  - Edges:  $(i, a) \in E \subseteq V \times F$
  - F(i)/V(a) = set of neighbors for node-i/a
  - Messages:  $\mu_{i \to a}^{(t)}(x_i)$  and  $\hat{\mu}_{a \to i}^{(t)}(x_i)$

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- ▶ variable-*i* to factor-*a* message



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- variable-i marginal


iteration 1: variable to factor

$$\mu_{i \to a}^{(1)}(x_i) = 1$$





iteration 1: factor to variable

$$\hat{\mu}_{4\to4}^{(1)}(x_4) = \sum_{x_5} f_4(x_4, x_5) \mu_{5\to4}^{(1)}(x_i)$$
$$= \sum_{x_5} f_4(x_4, x_5)$$
$$\hat{\mu}_{2\to4}^{(1)}(x_4) = f_3(x_4)$$

$$\mu_{4\to2}^{(2)}(x_4) = \hat{\mu}_{4\to4}^{(1)}(x_4)\hat{\mu}_{3\to4}^{(1)}(x_4)$$
$$= f_3(x_4)\sum_{x_5}f_4(x_4, x_5)$$

$$\mu_{6\to 2}^{(2)}(x_6) = 1$$



iteration 2: variable to factor 
$$\begin{split} \mu_{4\to 2}^{(2)}(x_4) &= \hat{\mu}_{4\to 4}^{(1)}(x_4)\hat{\mu}_{3\to 4}^{(1)}(x_4) \\ &= f_3(x_4)\sum f_4(x_4,x_5) \end{split}$$

$$\mu_{6\to 2}^{(2)}(x_6) = 1$$

iteration 2: factor to variable

$$\hat{\mu}_{2\to1}^{(2)}(x_1) = \sum_{x_4,x_6} f_2(x_1,x_4,x_6)\mu_{4\to2}^{(2)}(x_4)\mu_{6\to2}^{(2)}(x_6)$$
$$= \sum_{x_4,x_6} f_2(x_1,x_4,x_6)f_3(x_4)\sum_{x_5} f_4(x_4,x_5)$$
$$= f_2''(x_1)$$

 $x_{5}$ 

 $f_1$ 

 $x_2$ 

 $x_3$ 

 $f_2$ 

 $f_4$ 

 $x_6$ 

 $x_4$ 

fз

iteration 2: variable marginal

$$\mu_1^{(3)}(x_1) = \hat{\mu}_{1 \to 1}^{(2)}(x_1) \hat{\mu}_{2 \to 1}^{(2)}(x_1) = f_1''(x_1) f_2''(x_2)$$

Same answer as peeling but from a distributed parallel algorithm



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	2		5		1		9	
8			2		3			6
	3			6			7	
		1				6		
5	4						1	9
		2				7		
	9			3			8	
2			8		4			7
	1		9		7		6	

rows are permutations of  $\{1, 2, \ldots, 9\}$ 

	2		5		1		9	
8			2		3			6
	3			6			7	
		1				6		
5	4						1	9
		2				7		
	9			3			8	
2			8		4			7
	1		9		7		6	

rows are permutations of  $\{1, 2, \dots, 9\}$ columns are permutations of  $\{1, 2, \dots, 9\}$ 

	2		5		1		9	
8			2		3			6
	3			6			7	
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2			8		4			7
	1		9		7		6	

rows are permutations of  $\{1, 2, \ldots, 9\}$ columns are permutations of  $\{1, 2, \ldots, 9\}$ subblocks are permutations of  $\{1, 2, \ldots, 9\}$ 

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$x_{11}$	$x_{12}$	$x_{13}$	$x_{14}$	$x_{15}$	$x_{16}$	$x_{17}$	$x_{18}$	$x_{19}$
$x_{21}$	$x_{22}$	$x_{23}$	$x_{24}$	$x_{25}$	$x_{26}$	$x_{27}$	$x_{28}$	$x_{29}$
$x_{31}$	$x_{32}$	<i>x</i> <sub>33</sub>	$x_{34}$	$x_{35}$	$x_{36}$	$x_{37}$	$x_{38}$	$x_{39}$
$x_{41}$	$x_{42}$	$x_{43}$	$x_{44}$	$x_{45}$	$x_{46}$	$x_{47}$	$x_{48}$	$x_{49}$
$x_{51}$	$x_{52}$	$x_{53}$	$x_{54}$	$x_{55}$	$x_{56}$	$x_{57}$	$x_{58}$	$x_{59}$
$x_{61}$	$x_{62}$	$x_{63}$	$x_{64}$	$x_{65}$	$x_{66}$	$x_{67}$	$x_{68}$	$x_{69}$
$x_{71}$	$x_{72}$	<i>x</i> <sub>73</sub>	$x_{74}$	$x_{75}$	$x_{76}$	x77	$x_{78}$	$x_{79}$
$x_{81}$	$x_{82}$	<i>x</i> <sub>83</sub>	$x_{84}$	$x_{85}$	$x_{86}$	<i>x</i> <sub>87</sub>	$x_{88}$	$x_{89}$
$x_{91}$	$x_{92}$	$x_{93}$	$x_{94}$	$x_{95}$	$x_{96}$	$x_{97}$	$x_{98}$	<i>x</i> <sub>99</sub>

implied factor graph has 81 variable and 27 factor nodes

$$f(\underline{x}) = \left(\prod_{i=1}^{9} f_{\sigma}(x_{i*})\right) \left(\prod_{j=1}^{9} f_{\sigma}(x_{*j})\right) \left(\prod_{k=1}^{9} f_{\sigma}(x_{B(k)})\right) \prod_{(i,j)\in O} \mathbb{I}(x_{ij} = y_{ij})$$

### Solving Sudoku with a Factor Graph

- Consider any constraint satisfaction problem with observed entries
  - One can write  $f(\underline{x})$  as the product of indicator functions
  - ▶ Some factors force <u>x</u> to be valid (i.e., satisfy constraints)
  - Other factors force  $\underline{x}$  to be compatible with observed values
  - Summing over  $\underline{x}$  counts the # of valid compatible sequences

### Solving Sudoku with a Factor Graph

- Consider any constraint satisfaction problem with observed entries
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  - Other factors force  $\underline{x}$  to be compatible with observed values
  - Summing over  $\underline{x}$  counts the # of valid compatible sequences
- Low-complexity peeling solution
  - Set elements of  $\underline{x}$  one at a time
  - Each step looks for  $i \in [n]$  and  $x' \in \mathcal{X}$  such that:
    - For currently set variables,  $f(\underline{x}) = 0$  for all  $x_i \in \mathcal{X} \setminus x'$
  - Sudoku's unique solution implies that  $x_i = x'$  correct
  - Fix  $x_i = x'$  and repeat until all values fixed

### Boolean Satisfiability: K-SAT

▶ One instance of 3-SAT is given, for example, by

 $f(\underline{x}) = (\overline{x}_1 \lor \overline{x}_3 \lor x_7) \land (x_1 \lor \overline{x}_2 \lor x_5) \land (x_2 \lor \overline{x}_4 \lor x_6).$ 

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Marginalization allows uniform sampling from valid set

For i = 1, 2, ..., n, fix  $x_j$  for j < i and compute marginal

$$g_i(x_i) = \frac{1}{Z_i} \sum_{x_{i+1}, \dots, x_n} f(\underline{x}) = \mathbb{P}\left(X_i = x_i | X_1 = x_1, \dots, X_{i-1} = x_{i-1}\right)$$

• Then, sample  $x_i \sim g_i(\cdot)$  and repeat

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- This algorithm has low complexity if factor graph forms a tree
  - If not a tree, use approximate marginal from belief propagation
  - This is related to BP-guided decimation [MM09]

## Low-Density Parity-Check (LDPC) Codes



▶ Linear codes defined by  $\underline{x}H^T = \underline{0}$  for all c.w.  $\underline{x} \in \mathcal{C} \subset \{0,1\}^n$ 

- $\blacktriangleright~H$  is an  $m\times n$  sparse parity-check matrix for the code
- $\blacktriangleright$  Code bits and parity checks associated with cols/rows of H

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Factor graph: H is the biadjacency matrix for variable/factor nodes

- Ensemble defined by configuration model for random graphs
- Checks define factors:  $f_{\text{even}}(x_1^d) = \mathbb{I}(x_1 \oplus \cdots \oplus x_d = 0)$
- Let  $\underline{x}_{\partial a}$  be the subvector of variables in the *a*-th check and

$$f(x_1, \dots, x_n) = \left(\prod_{a=1}^m f_{\text{even}}(\underline{x}_{\partial a})\right) \left(\prod_{i=1}^n P_{Y|X}(y_i|x_i)\right)$$

## A Little History

### introduced LDPC codes in 1962 paper

1962

IRE TRANSACTIONS ON INFORMATION THEORY

Low-Density Parity-Check Codes\*

R. G. GALLAGER<sup>†</sup>

Summary—A low-density parity-check code is a code specified by a parity-check matrix with the following properties cash column contains a small fixed number  $j \geq 3$  of 1's and each row contains a small fixed number k > j of 1's and the typical minimum distance of these codes increases linearly with block length for a fixed rate and (inclusion) in the structure channel, the typical probability of decoding error decreases exponentially with block length for a fixed rate and fixed f.

A simple but nonoptimum decoding scheme operating directly from the channel a posteriori probabilities is described. Both the equations. We call the set of digits contained in a parity-check equation a parity-check set. For example, the first parity-check set in Fig. 1 is the set of digits (1, 2, 3, 5).

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The use of parity-sheek codes makes coding (as distinguished from decoding) relatively simple to implement. Also, as Elias [3] has shown, if a typical parity-check code of long block length is used on a binary symmetric channel, and if the code rate is between *critical rate* and channel eapacity, then the probability of decoding error



Robert Gallager

### Judea Pearl



### defined general belief-propagation in 1986 paper

# Fusion, Propagation, and Structuring in Belief Networks\*

#### Judea Pearl

Cognitive Systems Laboratory, Computer Science Department, University of California, Los Angeles, CA 90024, U.S.A.

Recommended by Patrick Hayes

#### ABSTRACT

Belig networks are directed acyclic graphs in which the nodes represent proposition (or variable), the arcs signif and ext dependencies to heneven the linked propositions, and the strength of these dependencies are quantified by conditional probabilities. A network of this zort can be used to represent the generic knowledge of a domain expert, and it turns into a compositional architecture if the links are used not meetly for storing facual knowledge but also for directing and activating the data flow in the comparison which manipulate thit knowledge.

- Constraint nodes define the valid patterns
  - Circles represent a single value shared by factors
  - Squares assert attached variables sum to 0 mod 2
- Iterative decoding on the binary erasure channel (BEC)
  - Messages passed in phases: bit-to-check and check-to-bit
  - Each output message depends on other input messages
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Capacity Achieving Codes: There and Back Again

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Capacity Achieving Codes: There and Back Again

# Density Evolution (DE) for LDPC Codes



- Binary erasure channel (BEC) with erasure prob.  $\varepsilon$
- ▶ DE tracks bit-to-check msg erasure rate  $x_\ell$  after  $\ell$  iterations
- $\blacktriangleright$  Defines noise threshold  $\varepsilon^{\rm BP}$  for the large system limit
  - Easily computed numerically for given code ensemble

Introduced by ten Brink in 1999 to understand iterative decoding

▶ For the BEC, the MAP EXIT curve is

$$h^{\mathrm{MAP}}(\varepsilon) \triangleq \frac{1}{n} \sum_{i=1}^{n} H(X_i | \underline{Y}_{\sim i}(\varepsilon))$$

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$$h^{\mathrm{MAP}}(\varepsilon) \triangleq \frac{1}{n} \sum_{i=1}^{n} H(X_i | \underline{Y}_{\sim i}(\varepsilon))$$

► EXIT Area Theorem [ABK04] $\frac{1}{n}H(\underline{X}|\underline{Y}(\varepsilon)) = \int_0^{\varepsilon} h^{\mathrm{MAP}}(\delta) \mathrm{d}\delta$ 

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- BP EXIT curve

$$h^{\mathrm{BP}}(\varepsilon) \triangleq \frac{1}{n} \sum_{i=1}^{n} H\left(X_i | \Phi_i^{\mathrm{BP}}(\underline{Y}_{\sim i}(\varepsilon))\right)$$

- where  $\Phi_i^{\mathrm{BP}}(Z)$  is the BP estimate of  $X_i$  given Z
- ▶ Data processing inequality:  $h^{\mathrm{BP}}(\varepsilon) \ge h^{\mathrm{MAP}}(\varepsilon)$



- ► (3,4)-regular LDPC code
  - Codeword  $(X_1, \ldots, X_n)$
  - Received  $(Y_1, \ldots, Y_n)$



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- MAP EXIT curve is extrinsic entropy  $H(X_i | \underline{Y}_{\sim i})$  vs. channel  $\varepsilon$ 
  - 0 below MAP threshold 0.746
  - Area under curve equals rate R
  - Upper bounded by BP EXIT
# EXtrinsic Information Transfer (EXIT) Curves



- (3,4)-regular LDPC code
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  - ► Area under curve equals rate *R*
  - Upper bounded by BP EXIT
- MAP threshold upper bound  $\overline{\varepsilon}^{MAP}$ 
  - $\varepsilon$  s.t. area under BP EXIT is R

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## Properties of the MAP EXIT Curve

▶ For linear codes, the recovery of  $X_i$  from  $\underline{Y} = y$ 

- $\blacktriangleright$  is independent of the transmitted codeword  $\underline{X}$
- only depends on erasure indicator  $z_i = \mathbf{1}_{\{?\}}(y_i)$
- ▶ is determined by whether  $H(X_i | \underline{Z} = \underline{z})$  is 0 or 1

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- The MAP bit-erasure rate  $P_b(\varepsilon)$  satisfies

$$P_b(\varepsilon) = \mathbb{P}(Y_i = ?)H(X_i | \underline{Y}, Y_i = ?) = \varepsilon h^{\text{MAP}}(\varepsilon)$$

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- ▶ A sequence of rate-*R* codes achieves capacity iff
  - $P_b(\varepsilon) \to 0$  for all  $\varepsilon < 1 R$
  - $\blacktriangleright \ h^{\mathrm{MAP}}(\varepsilon) \to 0 \text{ for all } \varepsilon < 1-R$
  - $h^{\mathrm{MAP}}(\varepsilon)$  transitions sharply from 0 to 1



▶ For  $\delta > 0$ , transition width is  $\varepsilon$ -range over which  $\delta \le h^{MAP}(\varepsilon) \le 1 - \delta$ 



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- When do EXIT curves have a sharp transition? [KKMPSU15]
  - If the code's permutation group is doubly transitive!
  - ▶ For example, Reed-Muller and prim. narrow-sense BCH codes

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# Summary and Open Problems

- Gallager's 1960 thesis already contains most of the tools necessary to achieve capacity in practice
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- The first deterministic sequence of capacity-achieving binary codes for the BEC (under MAP decoding) was defined in 1954!
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- The first deterministic sequence of capacity-achieving binary codes for the BEC (under MAP decoding) was defined in 1954!
  - Sequences of Reed-Muller codes achieve capacity on the BEC
  - But, we didn't know this until 2015!
- Open problems
  - Generalize the Reed-Muller result to have weaker conditions and/or apply to more general channels/problems
  - Find a purely information-theoretic proof of the Reed-Muller result for the BEC

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# What is Spatial Coupling?

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# What is Spatial Coupling?



- Variable nodes have a natural global orientation
- Boundaries help variables to be recovered in an ordered fashion









Historical Notes

- LDPC convolutional codes introduced by FZ in 1999
- ▶ Shown to have near optimal noise thresholds by LSZC in 2005
- $\blacktriangleright~(l,r,L,w)$  ensemble proven to achieve capacity by KRU in 2011

IEEE TRANSACTIONS ON INFORMATION THEORY, VOL. 56, NO. 10, OCTOBER 2010

# Iterative Decoding Threshold Analysis for LDPC Convolutional Codes

Michael Lentmaier, Member, IEEE, Arvind Sridharan, Member, IEEE, Daniel J. Costello, Jr., Life Fellow, IEEE, and Kamil Sh. Zigangirov, Fellow, IEEE



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IEEE TRANSACTIONS ON INFORMATION THEORY, VOL. 57, NO. 2, FEBRUARY 2011

# Threshold Saturation via Spatial Coupling: Why Convolutional LDPC Ensembles Perform So Well over the BEC

Shrinivas Kudekar, Member, IEEE, Thomas J. Richardson, Fellow, IEEE, and Rüdiger L. Urbanke



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## Properties of Threshold Saturation

l	r	$\varepsilon^{\mathrm{BP}}$	$\varepsilon^{\mathrm{MAP}}$
3	6	0.4294	0.4882
4	8	0.3834	0.4977
5	10	0.3416	0.4995
6	12	0.3075	0.4999
7	14	0.2798	0.5000

- Spatial coupling achieves the MAP threshold as  $w \to \infty$ 
  - BP threshold typically decreases after l = 3
  - MAP threshold is increasing in l, r for fixed rate
- Benefits and Drawbacks
  - ▶ For fixed *L*, minimum distance grows linearly with block length
  - Rate loss of O(w/L) is a big obstacle in practice

# Threshold Saturation via Spatial Coupling

- General Phenomenon (observed by Kudekar, Richardson, Urbanke)
  - BP threshold of the spatially-coupled system converges to the MAP threshold of the uncoupled system
  - Can be proven rigorously in many cases!

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  - Factor graph defines system of coupled particles
  - Valid sequences are ordered crystalline structures

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- Can be proven rigorously in many cases!
- Connection to statistical physics
  - Factor graph defines system of coupled particles
  - Valid sequences are ordered crystalline structures
- Between BP and MAP threshold, system acts as supercooled liquid
  - Correct answer (crystalline state) has minimum energy
  - Crystallization (i.e., decoding) does not occur without a seed
  - Ex.: ice melts at  $0 \,^{\circ}$ C but freezing w/o a seed requires  $-48.3 \,^{\circ}$ C

#### http://www.youtube.com/watch?v=Xe8vJrlvDQM

# Why is Spatial Coupling Interesting?

- Breakthroughs: first practical constructions of
  - universal codes for binary-input memoryless channels [KRU12]
  - information-theoretically optimal compressive sensing [DJM11]
  - universal codes for Slepian-Wolf and MAC problems [YJNP11]
  - codes  $\rightarrow$  capacity with iterative hard-decision decoding [JNP12]
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  - Our proof for increasing scalar/vector recursions [YJNP12/13]

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  - ► Original proofs [KRU11/12] quite specific to LDPC codes
  - Our proof for increasing scalar/vector recursions [YJNP12/13]
- Spatial coupling as a proof technique [GMU13]
  - ▶ For a large random factor graph, construct a coupled version
  - Use DE to analyze BP decoding of coupled system
  - Compare uncoupled MAP with coupled BP via interpolation

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## Universality over Unknown Parameters

- The Achievable Channel Parameter Region (ACPR)
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- Properties
  - ► For fixed encoders, the ACPR depends on the decoders
  - ▶ For example, one has  $BP-ACPR \subseteq MAP-ACPR$
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  - For example, one has  $BP-ACPR \subseteq MAP-ACPR$
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- Universality
  - A sequence of encoding/decoding schemes is called universal if: its ACPR equals the optimal ACPR
  - Channel parameters are assumed unknown at the transmitter
  - At the receiver, the channel parameters are easily estimated

## 2-User Binary-Input Gaussian Multiple Access Channel



- Fixed noise variance
- Real channel gains  $h_1$  and  $h_2$  not known at transmitter
- Each code has rate R
- MAC-ACPR denotes the information-theoretic optimal region

# A Little History: SC for Multiple-Access (MAC) Channels

- ▶ KK consider a binary-adder erasure channel (ISIT 2011)
  - SC exhibits threshold saturation for the joint decoder
- > YNPN consider the Gaussian MAC (ISIT/Allerton 2011)
  - SC exhibits threshold saturation for the joint decoder
  - For channel gains h<sub>1</sub>, h<sub>2</sub> unknown at transmitter, SC provides universality
- Others consider CDMA systems without coding
  - TTK show SC improves BP demod of standard CDMA
  - ST prove saturation for a SC protograph-style CDMA

# Spatially-Coupled Factor Graph for Joint Decoder



### Spatially-Coupled Factor Graph for Joint Decoder



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Capacity Achieving Codes: There and Back Again



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• Smooth increasing  $f : [0, 1] \rightarrow [0, 1]$ 



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- Discrete-time recursion

$$x^{(\ell+1)} = f(x^{(\ell)})$$



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$$U_{\rm s}(x)$$

$$U_{\rm s}(x) = \int_0^x \left( z - f(z) \right) \mathrm{d}z = \frac{x^2}{2} - F(x)$$





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Capacity Achieving Codes: There and Back Again

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Both  $\downarrow 0$  iff no fixed points in (0,1]



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# Coupled Monotone Recursion (1)

► Coupled recursion 
$$\underline{x}^{(\ell+1)} = T \underline{x}^{(\ell)}$$
 with  $\underline{x}^{(\ell)} = \left(x_0^{(\ell)}, x_1^{(\ell)}, \ldots\right)$  and  
 $T \underline{x} \triangleq A^\top \underline{f}(A \underline{x}),$ 

where  $[\underline{f}(\underline{x})]_i = f(x_i)$  and A averages w adjacent values

$$A = \frac{1}{w} \begin{bmatrix} 1 & 1 & \cdots & 1 & 0 & \cdots \\ 0 & 1 & 1 & \ddots & 1 & \ddots \\ \vdots & \ddots & \ddots & \ddots & \ddots & \ddots \\ \end{bmatrix}$$

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• Coupled potential: 
$$U_{c}(\underline{x}) = \frac{1}{2} \sum_{i=0}^{\infty} x_{i}^{2} - \sum_{i=0}^{\infty} F\left(\frac{1}{w} \sum_{j=0}^{w-1} x_{i+j}\right)$$

• Satisfies 
$$\nabla U_{c}(\underline{x}) = \underline{x} - A^{\top} \underline{f}(A\underline{x})$$

Danger: there be dragons infinities

# Coupled Monotone Recursion (2)

- Properties of T (note:  $\underline{x} \preceq \underline{y} \Leftrightarrow x_i \leq y_i$  for all i)
  - T is monotone:  $\underline{x} \preceq \underline{y}$  implies  $T\underline{x} \preceq T\underline{y}$
  - ▶ T preserves spatial order:  $x_{i+1} \ge x_i$  implies  $[T\underline{x}]_{i+1} \ge [T\underline{x}]_i$

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▶ For  $\underline{x}^{(0)} = \underline{1}$ , iterates  $x_i^{(\ell)}$  are decreasing in  $\ell$  and increasing in i

- ▶ Spatial limit exists:  $x_{\infty}^{(\ell)} = \lim_{i \to \infty} x_i^{(\ell)}$
- ▶ Iteration limit exists:  $x_i^{(\infty)} = \lim_{\ell \to \infty} x_i^{(\ell)}$
- ▶ Iteration limit satisfies fixed point:  $\underline{x}^{(\infty)} = T\underline{x}^{(\infty)}$
- Double limit satisfies fixed point:  $x_{\infty}^{(\infty)} = f(x_{\infty}^{(\infty)})$

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$$[S\underline{x}]_i = x_{i-1}$$
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### Intuition Behind Threshold Saturation

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  - ▶ If  $x_{i+1} \ge x_i$  for all i, then  $V_{\underline{x}}(t)$  well-defined for  $t \in [0,1]$
  - For t = 1, one gets a telescoping sum that shows

$$V_{\underline{x}}(1) \le -U_{\mathrm{s}}(x_{\infty})$$

### Theorem

### If f(0)=0 and f'(0)<1 (0 is stable f.p.) with $U_{\rm s}(x)>0$ for $x \in (0,1]$ , then $\exists w_0 < \infty$ such that $x_{\infty}^{(\infty)} = 0$ for all $w > w_0$ .

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If f(0)=0 and f'(0)<1 (0 is stable f.p.) with  $U_s(x)>0$  for  $x \in (0,1]$ , then  $\exists w_0 < \infty$  such that  $x_{\infty}^{(\infty)} = 0$  for all  $w > w_0$ .

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  - $\blacktriangleright$  Thus, we get a contradiction for sufficiently large w

# History of Threshold Saturation Proofs

- the BEC in 2010 [KRU11]
  - Established many properties and tools used by later approaches
- ▶ the Curie-Weiss model of physics in 2010 [HMU12]
- CDMA using a GA in 2011 [TTK12]
- CDMA with outer code via GA in 2011 [Tru12]
- compressive sensing using a GA in 2011 [DJM13]
- regular codes on BMS channels in 2012 [KRU13]
- increasing scalar and vector recursions in 2012 [YJNP14]
- irregular LDPC codes on BMS channels in 2012 [KYMP14]
- non-decreasing scalar recursions in 2012 [KRU15]
- non-binary LDPC codes on the BEC in 2014 [AG16]
- ▶ and more since 2014...

#### ► Factor Graphs

- Useful tool for modeling dependent random variables
- Low-complexity algorithms for approximate inference
- Density evolution can be used to analyze performance
- Spatial Coupling
  - Powerful technique for designing and understanding FGs.
  - Related to the statistical physics of supercooled liquids
  - Simple proof of threshold saturation for scalar recursions
- Interesting Open Problems
  - Code constructions that reduce the rate-loss due to termination
  - Compute the scaling exponent for SC codes
  - Finding new problems where SC provides benefits

# Thanks for your attention

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