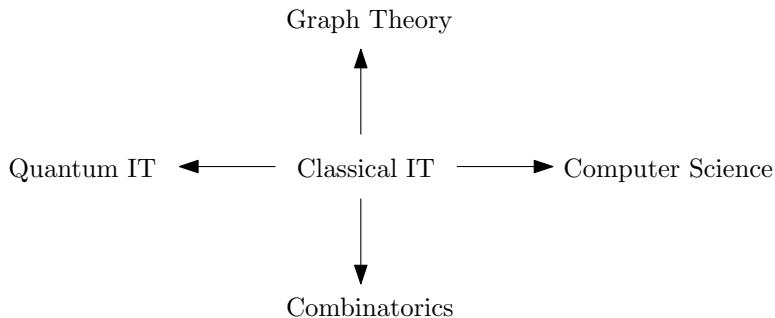


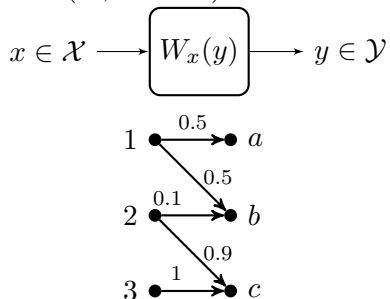
Channel reliability: from ordinary to zero-error capacity

Marco Dalai

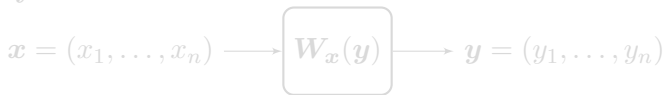
Department of Information Engineering
University of Brescia - Italy



- Discrete channel W (\mathcal{X}, \mathcal{Y} finite)

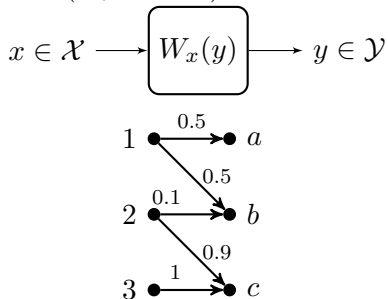


- Memoryless extension W

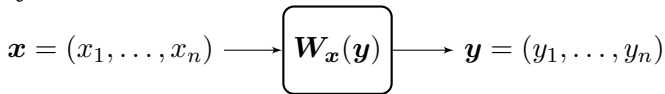


$$W_{\mathbf{x}}(\mathbf{y}) = \prod_i W_{x_i}(y_i)$$

- Discrete channel W (\mathcal{X}, \mathcal{Y} finite)



- Memoryless extension W



$$W_{\mathbf{x}}(\mathbf{y}) = \prod_i W_{x_i}(y_i)$$

- **Code:** M codewords $\{\mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_M\} \subset \mathcal{X}^n$
- **Decoder:** M disjoint decision regions $\{\mathcal{Y}_1, \dots, \mathcal{Y}_M\} \subseteq \mathcal{Y}^n$
(here: maximum likelihood decoder)
- **Probability of error** given message m

$$\begin{aligned} P_{e|m} &= \sum_{\mathbf{y} \notin \mathcal{Y}_m} W_{x_m}(\mathbf{y}) \\ &= W_{x_m}(\overline{\mathcal{Y}_m}) \end{aligned}$$

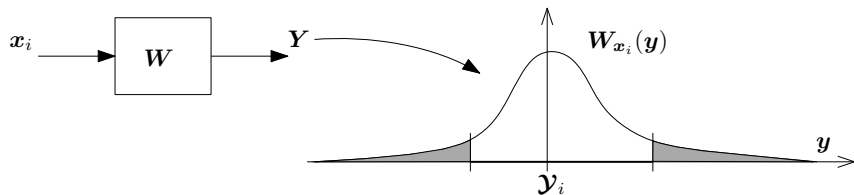
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Code and Error Probability

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- **Maximum error probability**

$$P_{e,\max} = \max_m P_{e|m}$$

- **Optimal codes**

$$P_{e,\max}^{(n)}(R) = \min_C P_{e,\max}$$

where the minimum is over codes of length n and rate at least R

- **Channel capacity**

$$C = \sup \left\{ R : \limsup_{n \rightarrow \infty} P_{e,\max}^{(n)}(R) = 0 \right\}$$

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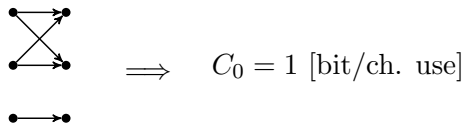
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- **Zero-error capacity**

$$C_0 = \sup\{R : P_{e,\max}^{(n)}(R) = 0 \text{ for some } n\}.$$

Example:



- **Reliability function:**

$$E(R) = \limsup_{n \rightarrow \infty} -\frac{1}{n} \log P_{e,\max}^{(n)}(R)$$

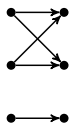
that is,

$$P_{e,\max}^{(n)}(R) \approx e^{-nE(R)} \quad C_0 < R < C$$

- Zero-error capacity

$$C_0 = \sup\{R : P_{e,\max}^{(n)}(R) = 0 \text{ for some } n\}.$$

Example:



$$\implies C_0 = 1 \text{ [bit/ch. use]}$$

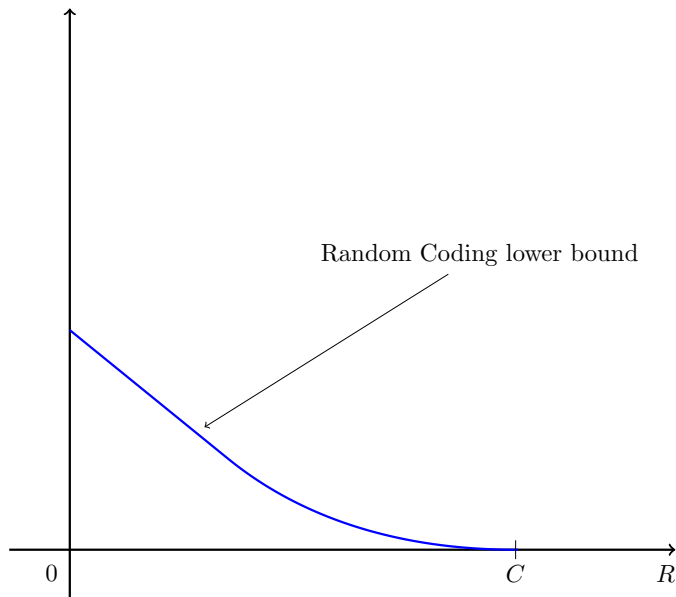
- Reliability function:

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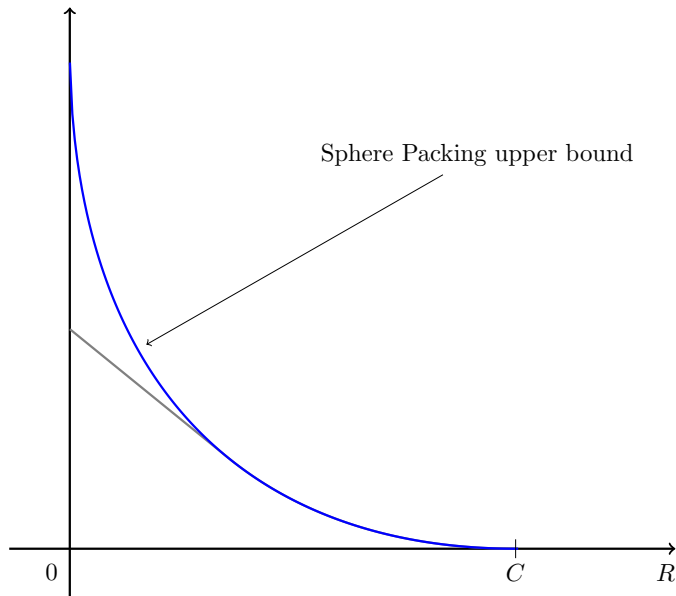
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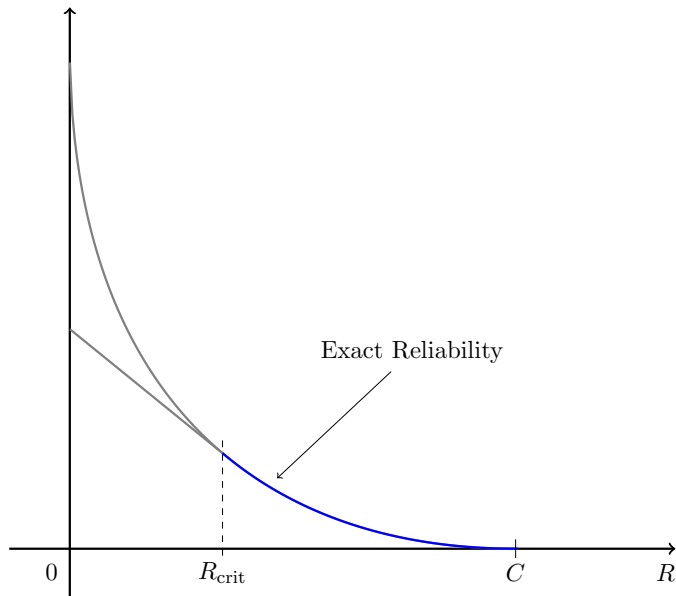
Bounds on $E(R)$: typical case with $C_0 = 0$



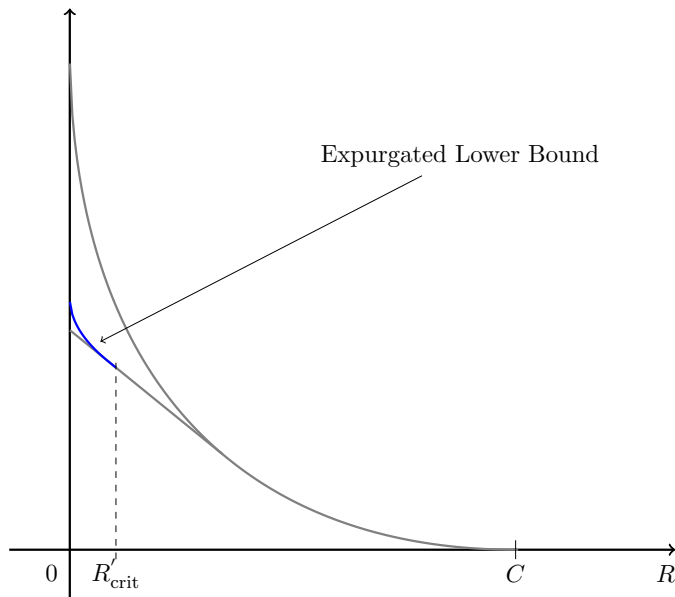
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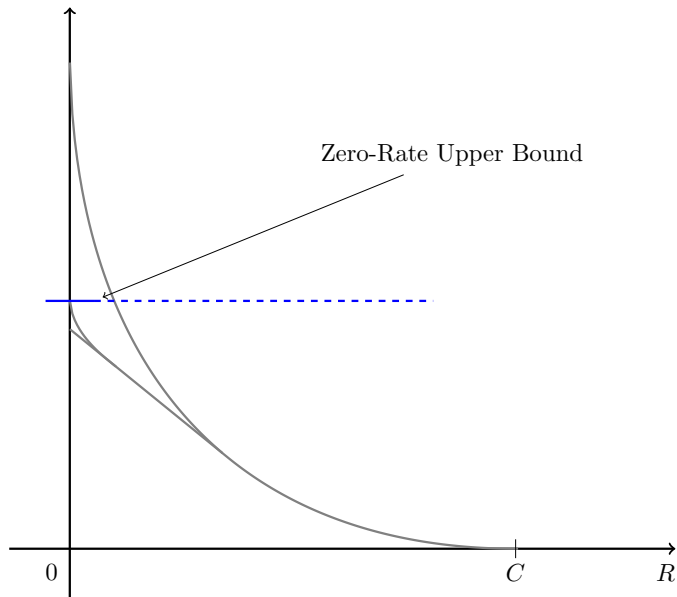
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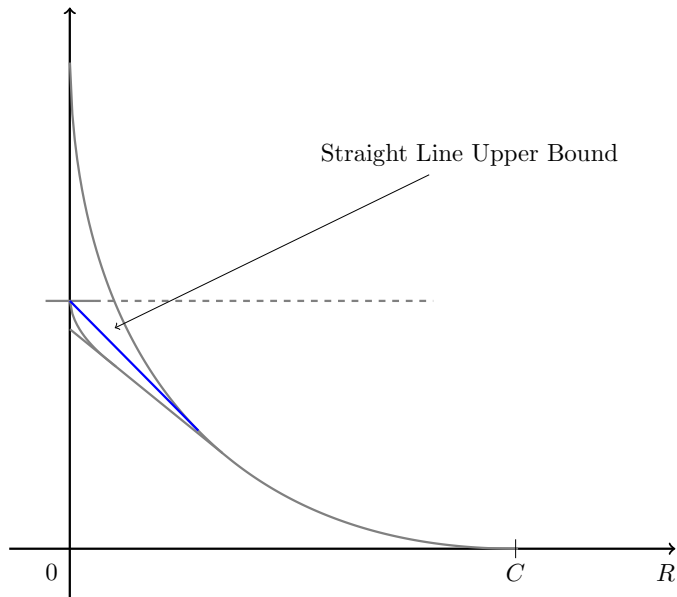
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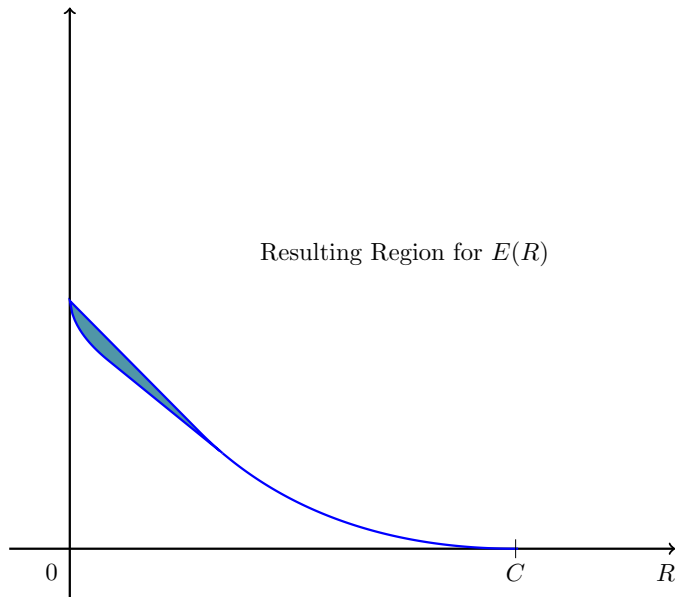
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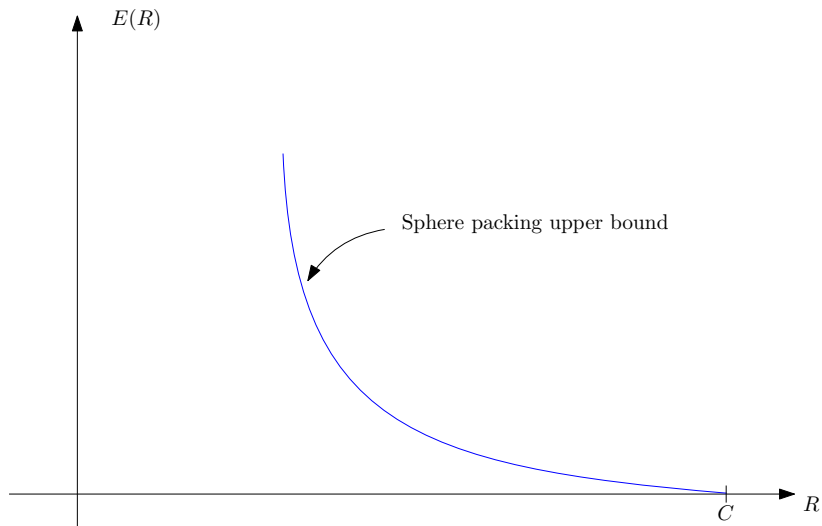
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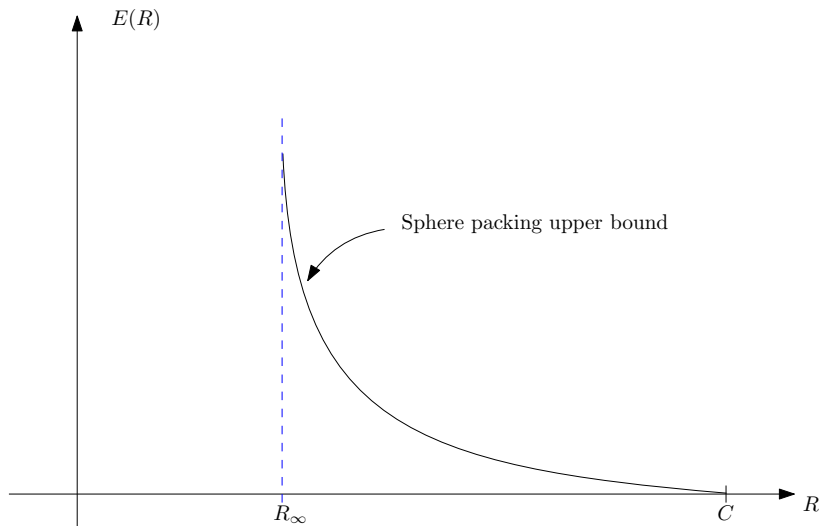
Bounds on $E(R)$: typical case with $C_0 = 0$



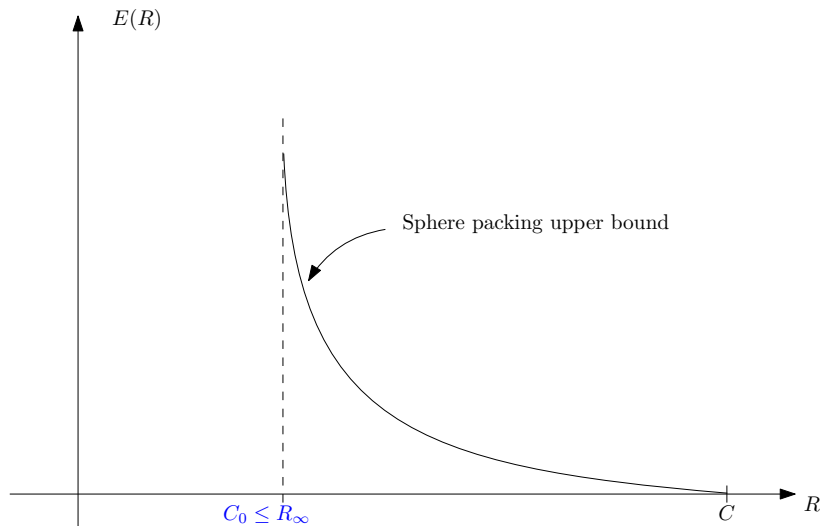
General case: $C_0 > 0$



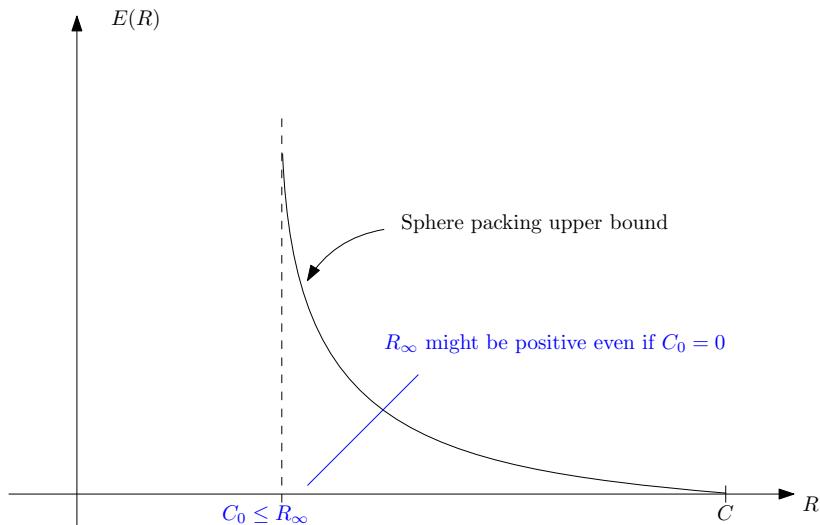
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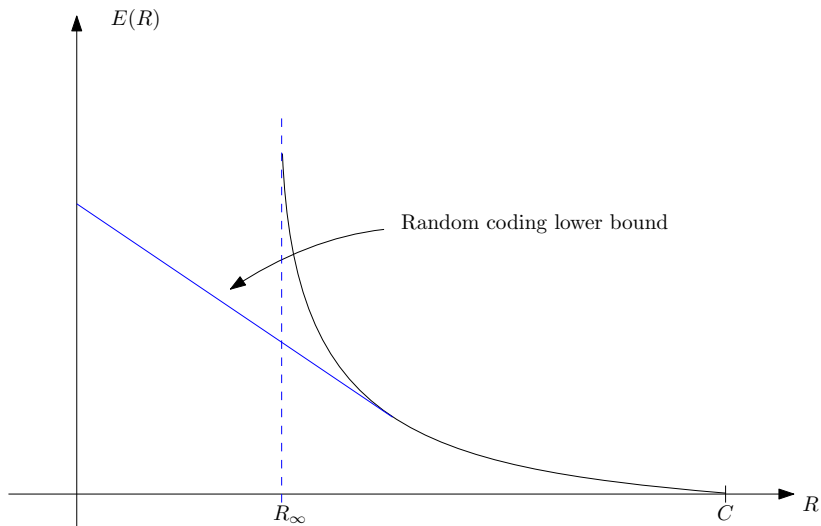
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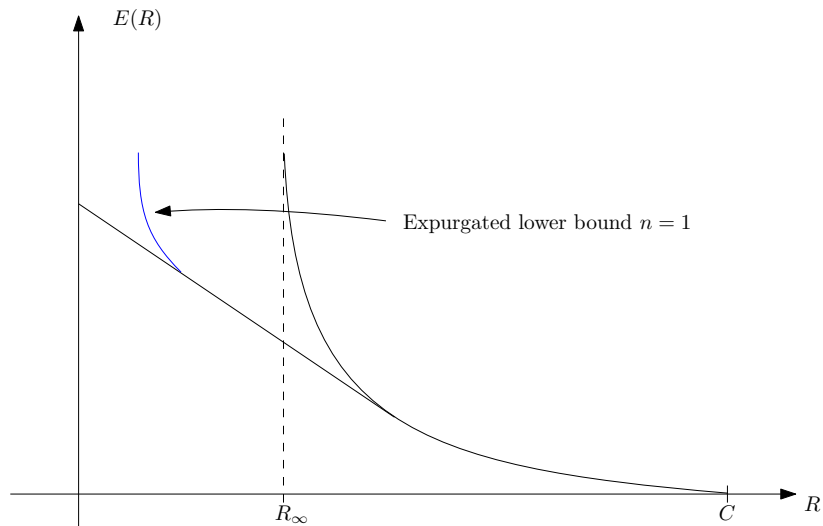
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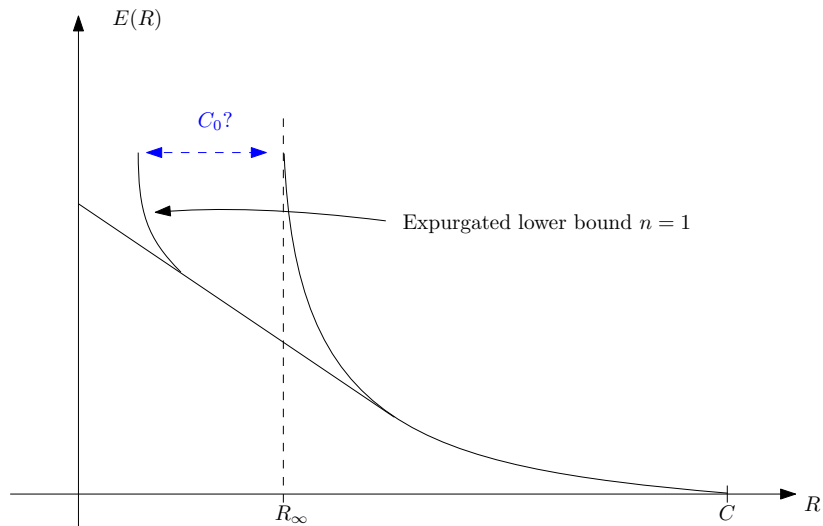
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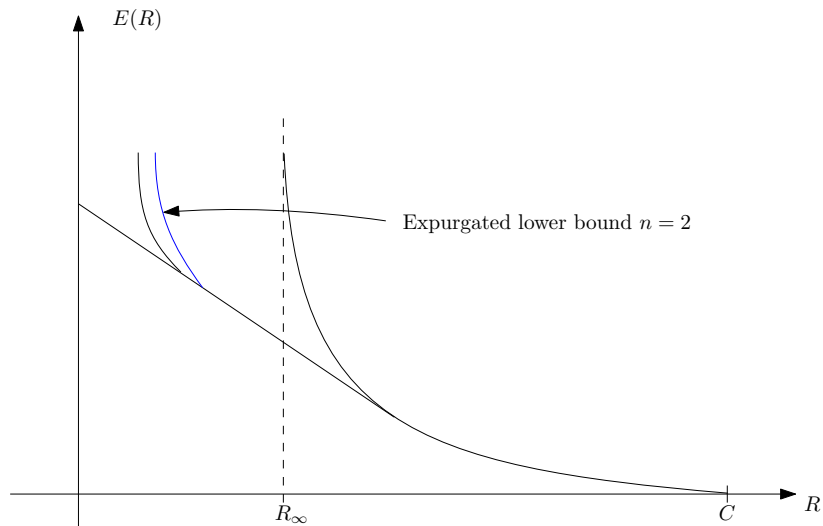
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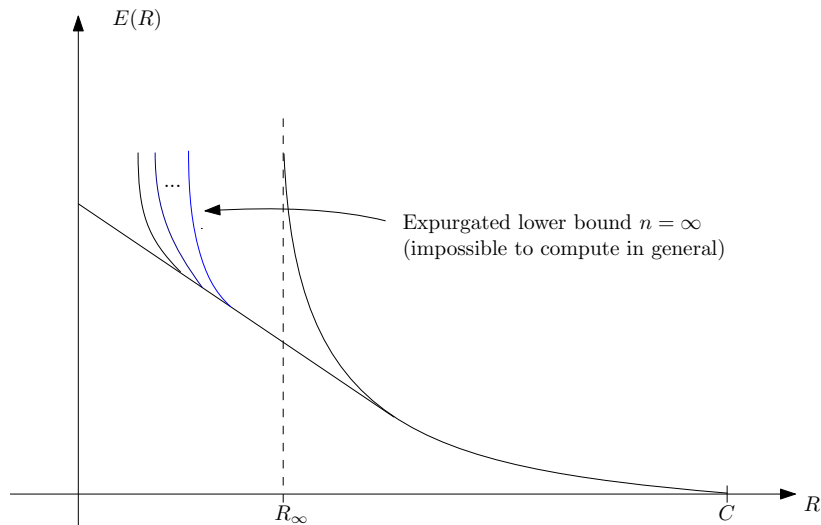
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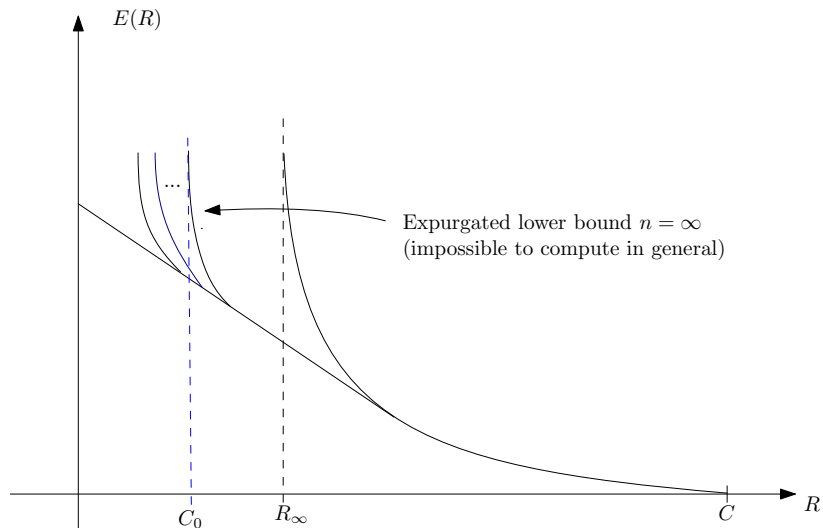
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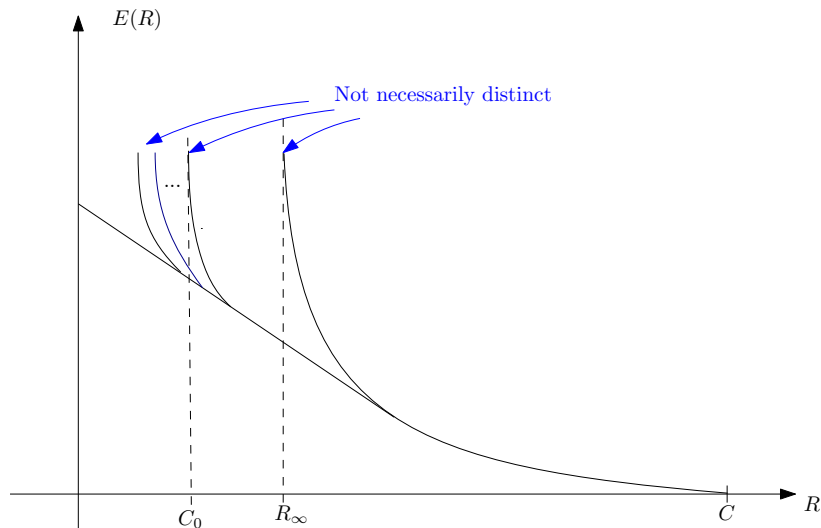
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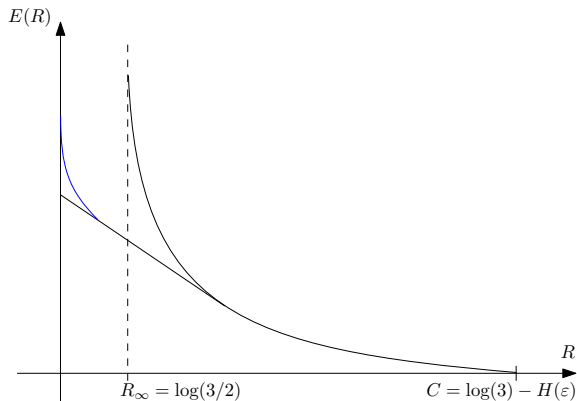
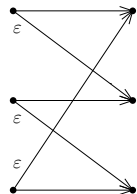
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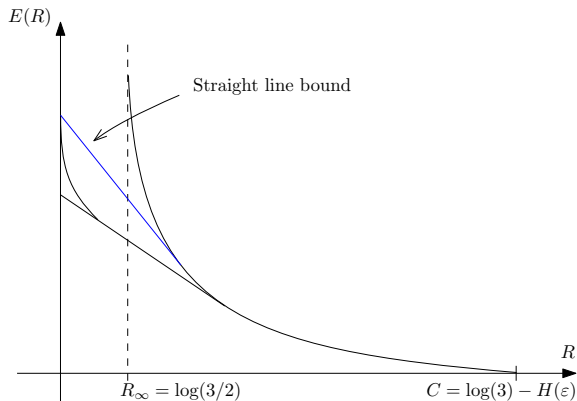
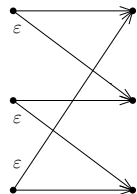


Example: typewriter channels



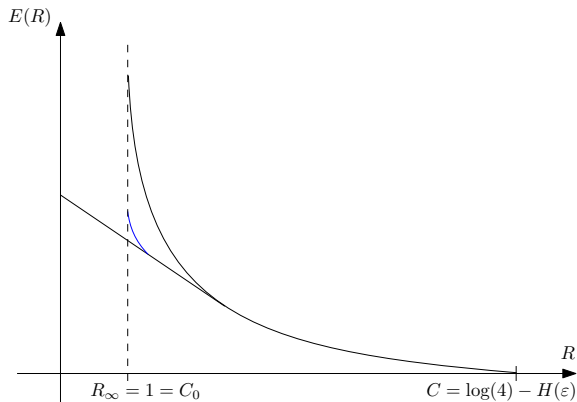
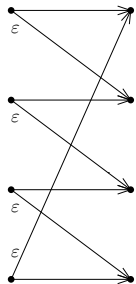
Note: $C_0 = 0$

Example: typewriter channels

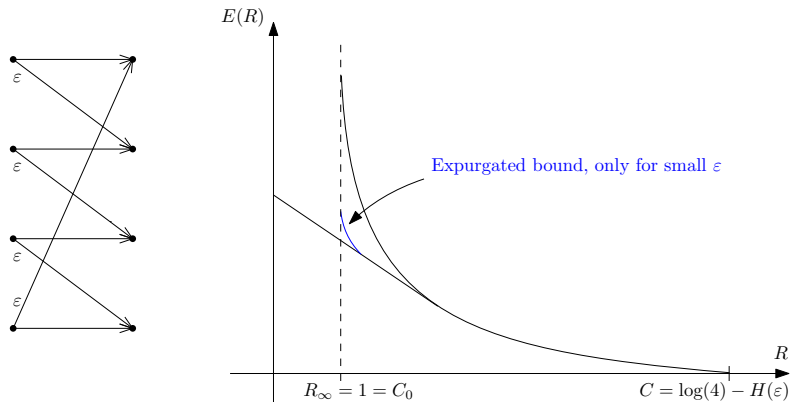


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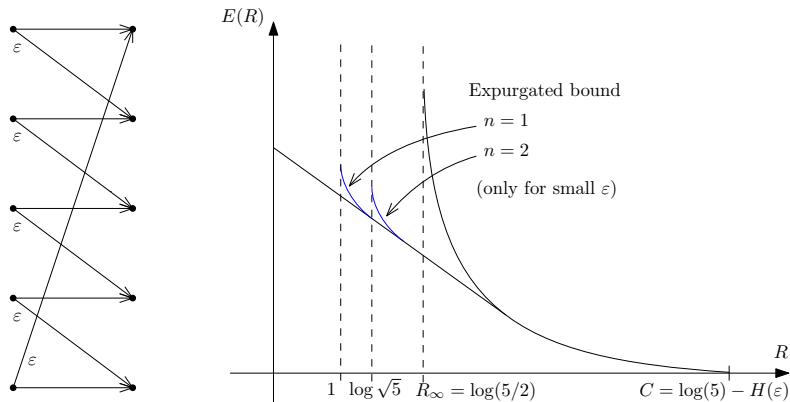
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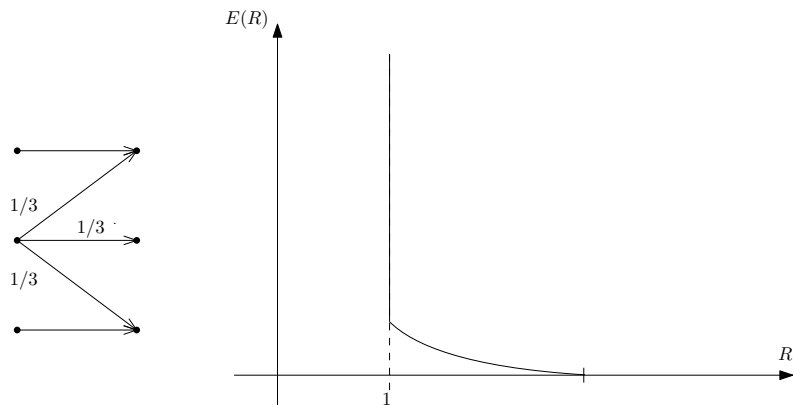


Example: typewriter channels



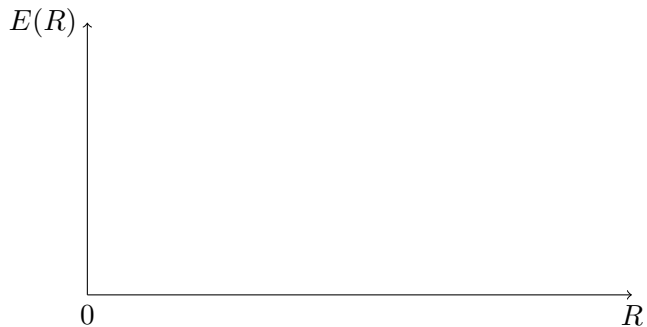
Note: $C_0 = \log \sqrt{5}$ (see later)

Example: strange case

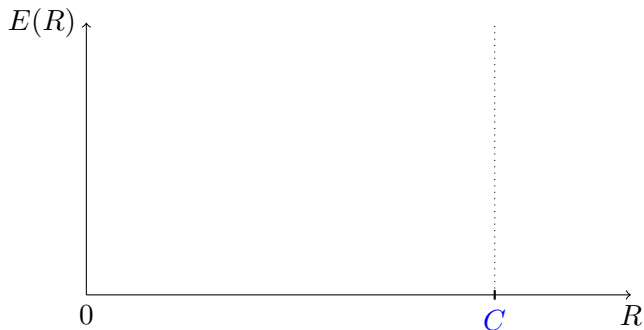


Note: exact reliability!

Some More Facts



Some More Facts



Shannon, 1948

$$C = \max_P \sum_{x,y} P(x) W_x(y) \log \frac{W_x(y)}{\sum_{x'} P(x') W_{x'}(y)},$$

Some More Facts

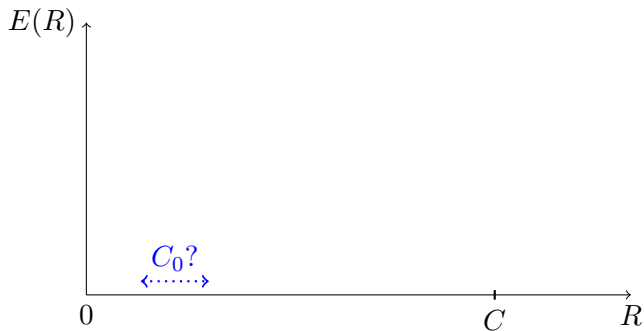


...later noticed to be an **information radius**

$$C = \min_Q \max_x D(W_x || Q)$$

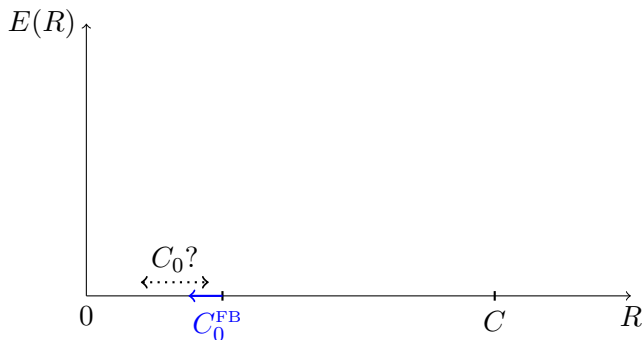
where $D(Q_1 || Q_2) = \sum_y Q_1(y) \log \frac{Q_1(y)}{Q_2(y)}$ is the Kullback-Leibler divergence

Some More Facts



Shannon, 1956

Some More Facts

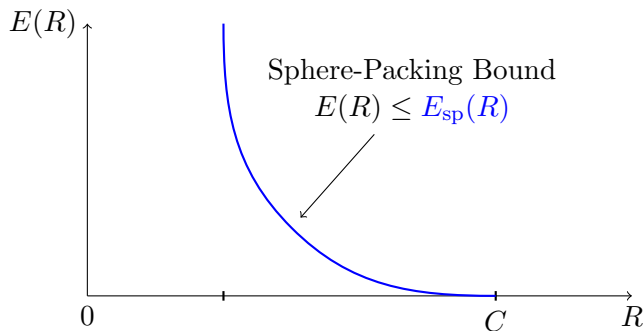


Shannon, 1956 (combinatorial)

Upper bounded by the zero-error capacity with feedback

$$C_0^{\text{FB}} = \max_P \left[-\log \max_y \sum_{x:W_x(y)>0} P(x) \right]$$

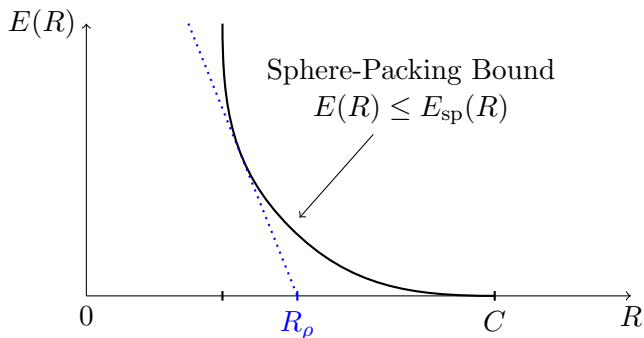
Some More Facts



Fano, 1961 - Shannon, Gallager and Berlekamp, 1967
(probabilistic)

$$E_{sp}(R) = \sup_{\rho \geq 0} \max_P \left[-\log \sum_y \left(\sum_x P(x) W_x(y)^{1/(1+\rho)} \right)^{1+\rho} - \rho R \right]$$

Some More Facts

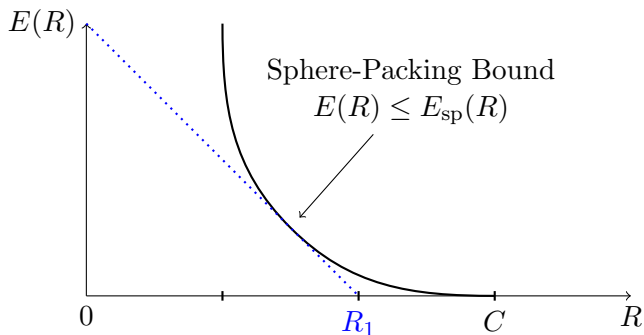


Also

$$R_\rho = \min_Q \max_x D_\alpha(W_x || Q), \quad \alpha = 1/(1 + \rho)$$

$D_\alpha(Q_1 || Q_2) = \frac{1}{\alpha-1} \log \sum_y Q_1(y)^\alpha Q_2(y)^{1-\alpha}$ is the Rényi divergence

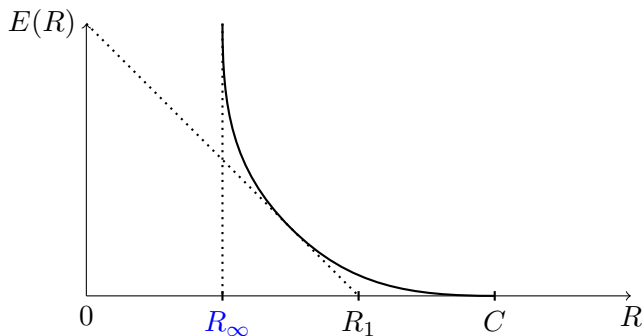
Some More Facts



Cutoff rate

$$R_1 = \min_Q \max_x \log \frac{1}{\left(\sum_y \sqrt{W_x(y)Q(y)} \right)^2}$$

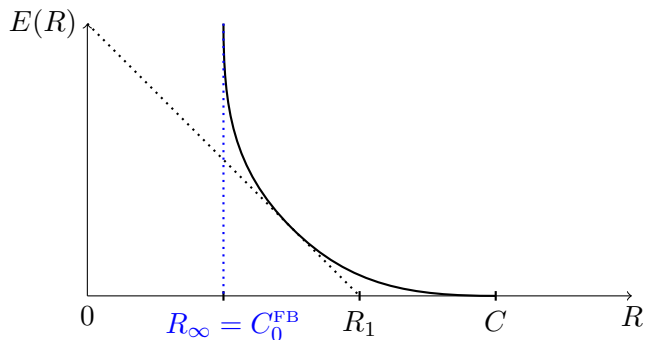
Some More Facts



R_∞ rate

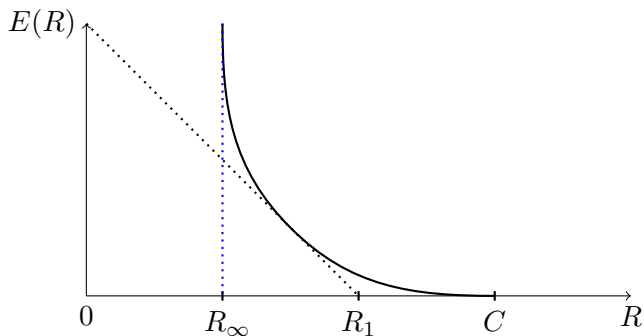
$$R_\infty = \min_Q \max_x \log \frac{1}{\sum_{y: W_x(y) > 0} Q(y)}$$

Some More Facts



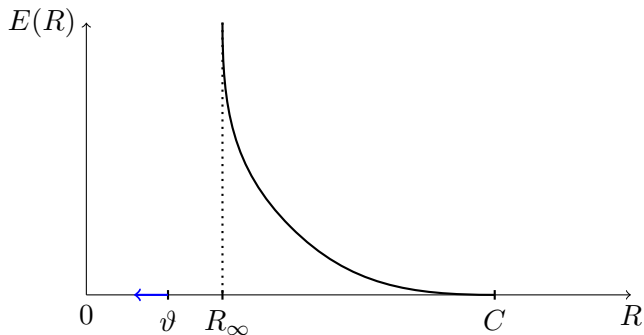
- $E_{\text{sp}}(R)$ gives $C_0 \leq R_\infty$
- So we have both $C_0 \leq C_0^{\text{FB}}$ and $C_0 \leq R_\infty$

Some More Facts



- $E_{\text{sp}}(R)$ gives $C_0 \leq R_\infty$
- So we have both $C_0 \leq C_0^{\text{FB}}$ and $C_0 \leq R_\infty$
- It turns out that $R_\infty = C_0^{\text{FB}}$ (whenever $C_0 > 0$)
- Same bound for C_0 using combinatorial or probabilistic approaches
- We can then minimize R_∞ over auxiliary channels \tilde{W}

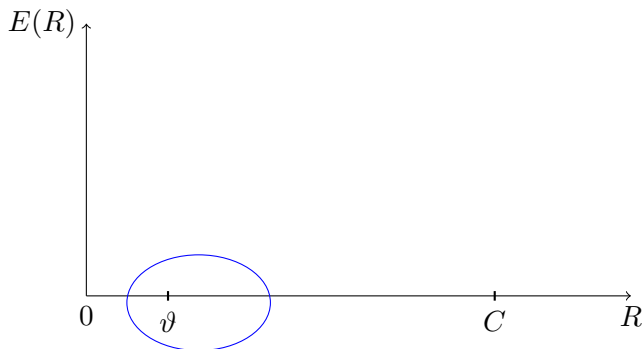
Some More Facts



Lovász, 1979

- New bound: $C_0 \leq \vartheta$
- Using *geometric representations of graphs*
- Combinatorial, apparently no connection with probability

Some More Facts

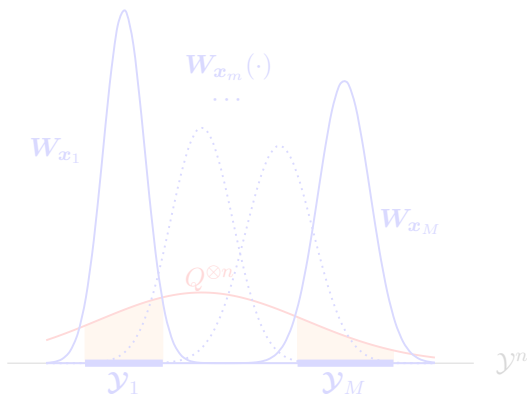


Lovász, 1979

- New bound: $C_0 \leq \vartheta$
- Using *geometric representations of graphs*
- Combinatorial, apparently no connection with probability
- **Goal:** better understanding of the R_∞ vs ϑ

Sphere-Packing Bound: Sketch of Proof

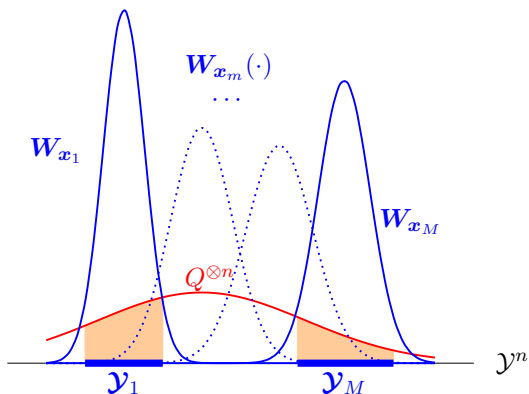
Binary hypothesis testing: compare $Q^{\otimes n}$ with W_{x_m}



- The decision regions $\mathcal{Y}_1, \dots, \mathcal{Y}_M$ are disjoint
- $Q^{\otimes n}(\mathcal{Y}_m) \leq 1/M$ for at least one m , since $\int Q^{\otimes n} = 1$
- $W_{x_m}(\overline{\mathcal{Y}_m}) \geq e^{-n(E_{sp}(R)+o(1))}$ using Neyman-Pearson/Chernoff

Sphere-Packing Bound: Sketch of Proof

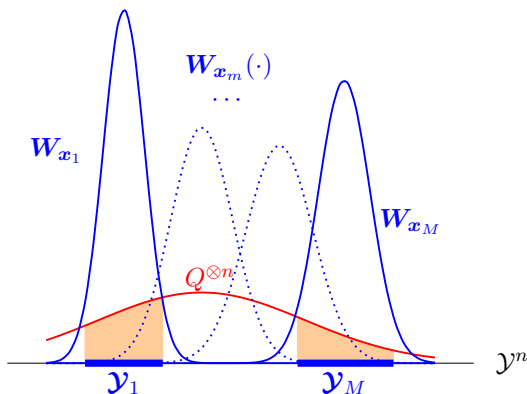
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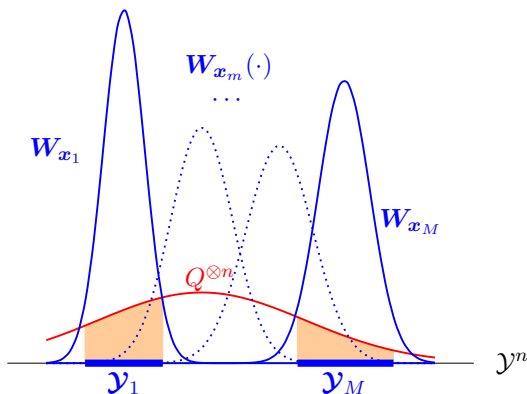
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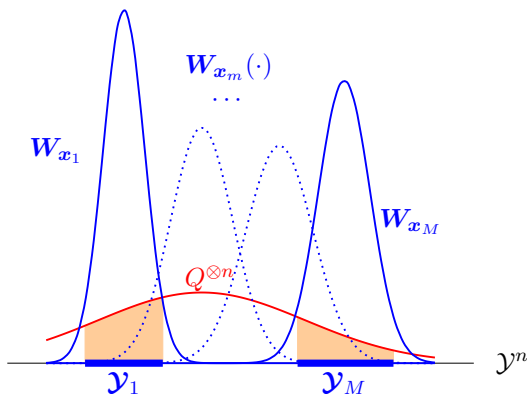
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Binary Hypothesis Testing (BHT)

BHT between distributions P_0 and P_1 over \mathcal{V} from n i.i.d. samples

- Two decision regions

\mathcal{V}_0 decision region
for P_0

P_0
•

\mathcal{V}_1 decision region
for P_1

P_1
•

- Error probabilities

$$P_{e|0} = \sum_{v \in \mathcal{V}_1} P_0(v),$$

$$P_{e|1} = \sum_{v \in \mathcal{V}_0} P_1(v)$$

Binary Hypothesis Testing (Rényi form)

Error exponents in BHT between P_0 and P_1 with n i.i.d. samples

$$\frac{1}{n} \log P_{e|0} = \mu(s) - s\mu'(s) + o(1)$$

$$\frac{1}{n} \log P_{e|1} = \mu(s) + (1-s)\mu'(s) + o(1)$$

where $0 < s < 1$,

$$\begin{aligned} \mu(s) &= \log \sum_{v \in \mathcal{V}} P_0(v)^{1-s} P_1(v)^s \\ &= -s D_{1-s}(P_0 \| P_1) \end{aligned}$$

and $D_\alpha(P \| Q)$ is the Rényi divergence

$$D_\alpha(P \| Q) = \frac{1}{\alpha - 1} \log \sum_{v \in \mathcal{V}} P^\alpha(v) Q^{1-\alpha}(v)$$

Note:

$$\lim_{\alpha \rightarrow 1} D_\alpha(P \| Q) = \sum_{v \in \mathcal{V}} P(v) \log \frac{P(v)}{Q(v)} =: D_{\text{KL}}(P \| Q)$$

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where $0 < s < 1$,

$$\begin{aligned} \mu(s) &= \log \sum_{v \in \mathcal{V}} P_0(v)^{1-s} P_1(v)^s \\ &= -sD_{1-s}(P_0 \| P_1) \end{aligned}$$

and $D_\alpha(P \| Q)$ is the Rényi divergence

$$D_\alpha(P \| Q) = \frac{1}{\alpha - 1} \log \sum_{v \in \mathcal{V}} P^\alpha(v) Q^{1-\alpha}(v)$$

Note:

$$\lim_{\alpha \rightarrow 1} D_\alpha(P \| Q) = \sum_{v \in \mathcal{V}} P(v) \log \frac{P(v)}{Q(v)} =: D_{\text{KL}}(P \| Q)$$

Binary Hypothesis Testing (Rényi form)

Error exponents in BHT between P_0 and P_1 with n i.i.d. samples

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Interpretation: Shannon-Gallager-Berlekamp, 1967

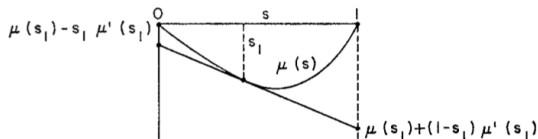
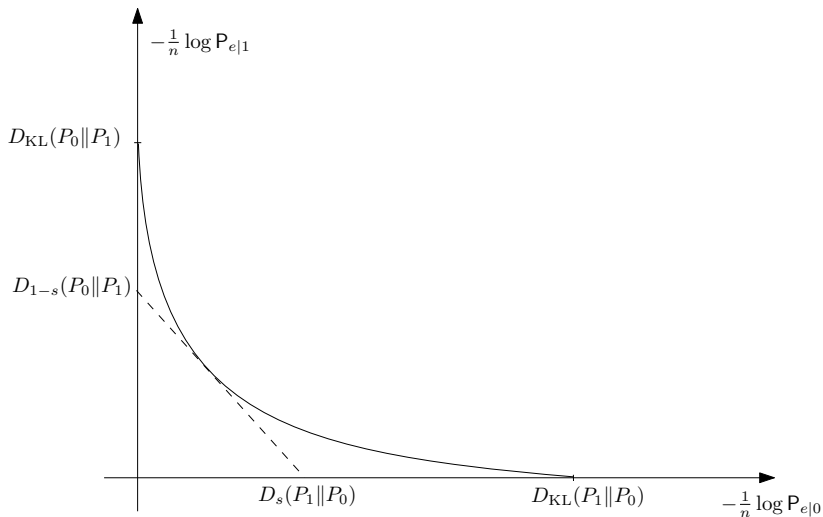


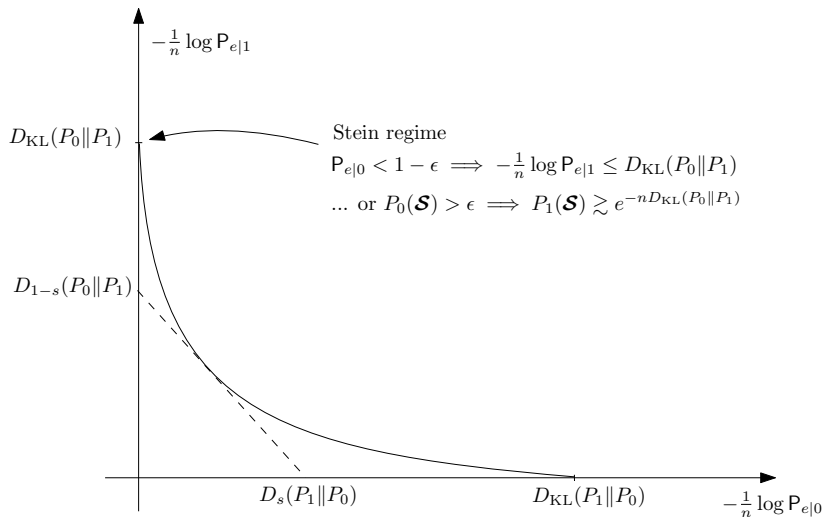
FIG. 6. Geometric interpretation of the exponents $\mu(s) - s\mu'(s)$ and $\mu(s) + (1 - s)\mu'(s)$.

Figure 6 gives a graphical interpretation of the terms $\mu(s) - s\mu'(s)$ and $\mu(s) + (1 - s)\mu'(s)$. It is seen that they are the endpoints, at 0 and 1, of the tangent at s to the curve $\mu(s)$. As s increases, the tangent see-saws around, decreasing $\mu(s) - s\mu'(s)$ and increasing $\mu(s) + (1 - s)\mu'(s)$. In the special case where $\mu(s)$ is a straight line, of course, this see-sawing does not occur and $\mu(s) - s\mu'(s)$ and $\mu(s) + (1 - s)\mu'(s)$ do not vary with s .

Another graphical representation



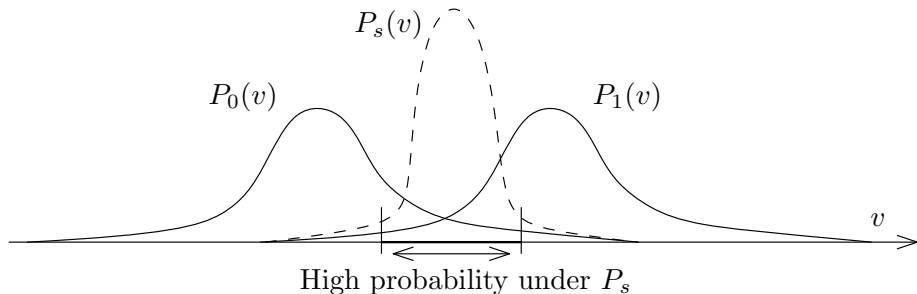
Another graphical representation



Binary Hypothesis Testing (Rényi form)

Key role played by the tilted mixture P_s

$$P_s(v) = \frac{P_0(v)^{1-s} P_1(v)^s}{\sum_{v'} P_0(v')^{1-s} P_1(v')^s} \implies \frac{P_0(v)}{P_s(v)} = e^{\mu(s)} e^{-s \log \frac{P_1(v)}{P_0(v)}}.$$



$$\frac{1}{n} \log \frac{P_1(\mathbf{v})}{P_0(\mathbf{v})} \approx \mu'(s) = \mathbb{E}_{P_s} \left[\log \frac{P_1(V)}{P_0(V)} \right]$$

Alternative expression (more popular)

$$-\frac{1}{n} \log P_{e|0} = D_{\text{KL}}(P_s \| P_0) + o(1)$$
$$-\frac{1}{n} \log P_{e|1} = D_{\text{KL}}(P_s \| P_1) + o(1)$$

- Very simple and intuitive: probabilities that P_0 and P_1 generate P_s -like sequences
- Directly uses the Stein regime in P_0 vs P_s and P_1 vs P_s
- Note (for later): this does *not* work in the quantum setting

Standard procedure for DMCs

- Given code with $M = e^{nR}$ codewords
- Group codewords by empirical “compositions” (or “type”, empirical frequency of symbols in the codeword)
- At most $n^{|\mathcal{X}|} = e^{o(n)}$ groups
- At least one group contains $e^{n(R-o(1))}$ codewords
- Bound probability of error for this subcode
- So, we can assume all codewords have same composition, say P

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Back to sphere packing: MIT Proof

- BHT between output distribution W_{x_m} and auxiliary $Q = Q^{\otimes n}$
- Use \mathcal{Y}_m as decision region for W_{x_m}
- $M = e^{nR}$ codewords; for at least one m , $Q(\mathcal{Y}_m) \leq 1/M$ and so

$$-\frac{1}{n} \log P_{e|Q} \geq R$$

- But for the optimal test

$$\begin{aligned} -\frac{1}{n} \log P_{e|W_{x_m}} &= -\mu(s) + s\mu'(s) + o(1) \\ -\frac{1}{n} \log P_{e|Q} &= -\mu(s) - (1-s)\mu'(s) + o(1) \end{aligned}$$

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So,

$$-\frac{1}{n} \log P_{e|W_{x_m}} \leq \sup_{0 < s < 1} \left[E_0(s, P) - \frac{s}{1-s} (R - \epsilon) \right] + o(1)$$

where

$$\begin{aligned} E_0(s, P) &= \min_Q \left[\frac{1}{s-1} \sum_x P(x) \log \sum_y W_x(y)^{1-s} Q(y)^s \right] \\ &= \min_Q \left[\frac{s}{1-s} \sum_x P(x) D_{1-s}(W_x \| Q) \right] \\ &= \frac{s}{1-s} I_{1-s}(P, W), \end{aligned}$$

where $I_\alpha(P, W)$ is Csiszár's version of α -mutual information.

The optimal Q is such that

$$Q(y) = \sum_x P(x) V_x(y)$$

if we define $V_x(y)$ as

$$V_x(y) = \frac{W_x^{1-s}(y) Q^s(y)}{\sum_{y'} W_x^{1-s}(y') Q^s(y')}.$$

This channel V is such that

$$\begin{aligned} I(P, V) &= \sum_x P(x) D(V_x \| Q) \\ &= R - \epsilon \end{aligned}$$

Sphere packing: Haroutunian's proof

- Consider an auxiliary channel V such that $I(P, V) < R$
- Converse: original coding scheme incurs $P_e > \epsilon$ on V
- For at least one codeword m , $V_{x_m}(\overline{\mathcal{Y}}_m) > \epsilon$.
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$$W_{x_m}(\overline{\mathcal{Y}}_m) \gtrsim e^{-nD(V\|W|P)}$$

- Optimizing over V

$$\frac{1}{n} \log \frac{1}{P_{e|W_{x_m}}} \leq \inf_{V: I(P, V) < R} D(V\|W|P)(1 + o(1)).$$

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- Just a single Q and M decoding regions implies $Q(\mathcal{Y}_m) \leq 1/M$ for some m
- If $Q(\mathcal{Y}_m) \leq e^{-nR}$ then $W_{x_m}(\overline{\mathcal{Y}_m})$ is at least $e^{-nE_{\text{sp}}(R)}$

Haroutunian

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Equivalent

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Zero-Error Capacity

- The zero-error capacity only depends on the *confusability* of symbols in the input alphabet \mathcal{X}
- Symbols x and x' confusable if $\exists y : W_x(y)W_{x'}(y) > 0$, or

$$\sum_y W_x(y)W_{x'}(y) > 0$$

- **Confusability graph**



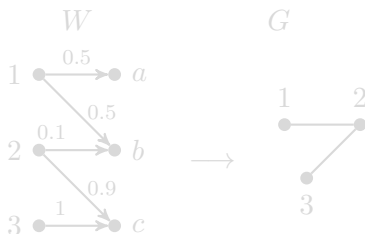
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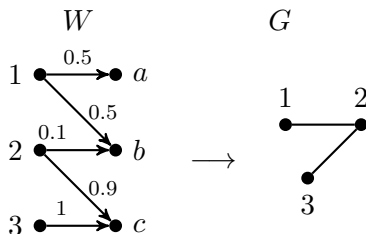
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Hence $C_0(W) = C(G)$ (*Graph Capacity*)

- Graph G
 - vertex set $V(G)$ (channel input symbols),
 - edge set $E(G)$ (pairs of distinct confusable symbols).
- $A \subseteq V(G)$ independent set if

$$x, x' \in A \implies x \not\sim x'$$

- Independence number

$$\alpha(G) = \max\{|A| : A \subseteq V(G) \text{ independent set}\}$$

- Strong power G^n
 - $V(G^n) = V(G) \times V(G) \cdots \times V(G) = V(G)^n$
 - $\mathbf{x} \neq \mathbf{x}'$ connected in G^n if entrywise either equal or connected in G

$$(x_1, x_2, \dots, x_n) \sim (x'_1, x'_2, \dots, x'_n) \iff \forall i, x_i \sim x'_i \text{ or } x_i = x'_i$$

i.e., confusable sequences.

So,

- $\alpha(G^n)$ is the largest size of an independent set in G^n or
- $\alpha(G^n)$ is the largest size of a zero-error code.

Graph Capacity

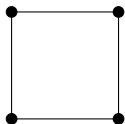
$$C(G) := \lim_{n \rightarrow \infty} \frac{1}{n} \log \alpha(G^n)$$

- $C(G)$ is highest asymptotic rate achievable with zero-error codes.
- Note: the limit exists due to Fekete's lemma since

$$\alpha(G^{n+m}) \geq \alpha(G^n)\alpha(G^m)$$

Three meaningful examples

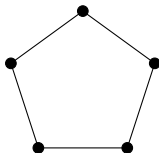
Square



$$C(G) = 2$$

(Shannon '56)

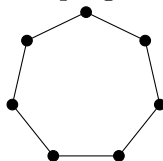
Pentagon



$$C(G) = \log \sqrt{5}$$

(Lovász '79)

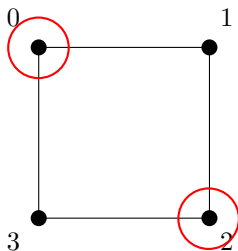
Heptagon



$$C(G)$$

unknown

Square: easy case



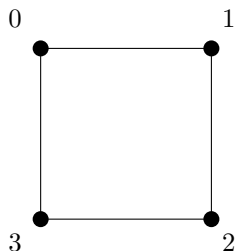
- Achievability:

$$\alpha(G) = 2 \implies \alpha(G^n) \geq 2^n \implies C(G) \geq 1 \text{ bit/ch. use}$$

- Converse

- Each sequence symbol either in A or in B
- 2^n “classes” of codewords
- Codewords in each class are all confusable.
- Pigeonhole principle: $\alpha(G^n) \leq 2^n$, so $C(G) \leq 1$

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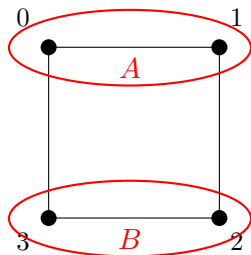
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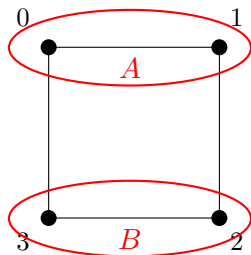
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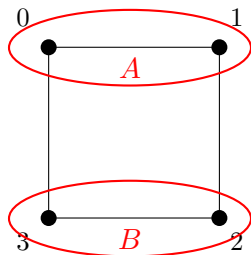
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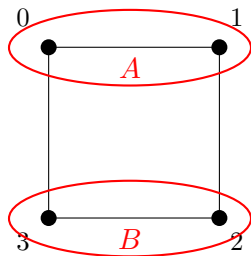
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[i+-i] Using the same reasoning

- Clique: subset of $V(G)$ completely connected in G (independent set in \bar{G})
- Assume G can be *covered* with k cliques
- Then $\alpha(G^n) \leq k^n$, and $C(G) \leq \log(k)$

Theorem

$$C(G) \leq \log \bar{\chi}(G)$$

where

$$\begin{aligned}\bar{\chi}(G) &= \text{clique covering number of } G \\ &= \text{minimum number of cliques to cover } G \\ &= \text{chromatic number of } \bar{G} \\ &=: \chi(\bar{G})\end{aligned}$$

Extension to fractional covers

- A set of cliques $A_1, \dots, A_k \subseteq V(G)$ is a fractional cover of G with weights $\lambda_1, \lambda_2, \dots, \lambda_k$ if

$$\sum_{i:v \in A_i} \lambda_i \geq 1, \quad \forall v \in V(G)$$

- Fractional clique covering number

$$\bar{\chi}^*(G) = \min \sum_i \lambda_i$$

minimum over fractional clique covers ($\lambda_1, \lambda_2, \dots, \lambda_k =$ weights).

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- Let $\lambda_1, \lambda_2, \dots, \lambda_k$ achieve $\bar{\chi}^*(G) = \sum_i \lambda_i$.
Define a probability distribution q on cliques

$$q_i = \frac{\lambda_i}{\sum_j \lambda_j}$$

- If A is random clique $\sim q$ then

$$P[v \in A] \geq \frac{1}{\sum_i \lambda_i}$$

- Pick random clique A in G^m as cartesian product of i.i.d. $\sim q$ cliques. Then,

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- So

$$E[|\mathcal{C} \cap \mathbf{A}|] \geq |\mathcal{C}| \cdot \left(\sum_i \lambda_i \right)^{-n}$$

- If \mathcal{C} is a zero-error code

$$\begin{aligned} 1 &\geq \max_{A \text{ clique}} |\mathcal{C} \cap A| \\ &\geq E[|\mathcal{C} \cap \mathbf{A}|] \\ &\geq |\mathcal{C}| \cdot \left(\sum_i \lambda_i \right)^{-n} \end{aligned}$$

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$$\begin{aligned} \alpha(G^n) &\leq \left(\sum_i \lambda_i \right)^n \\ &= \bar{\chi}^*(G)^n \end{aligned}$$

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- Setting $q(y) = \max_x \frac{Q(y)}{\sum_{y' \in \mathcal{Y}_x} Q(y')}$

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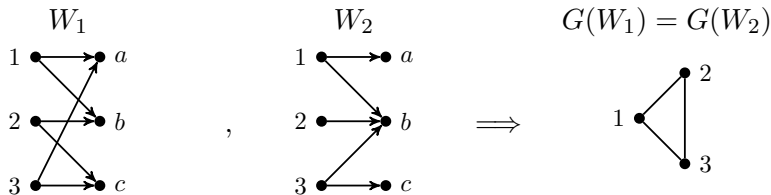
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Comparison with R_∞

- Indeed R_∞ does not only depend on $G(W)$

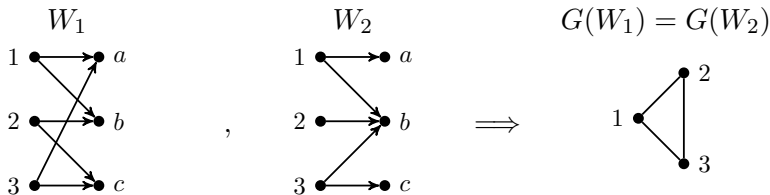


$$R_\infty(W_1) = \log 3/2 \quad R_\infty(W_2) = 0 \quad \text{but} \quad C(G) = 0$$

- To bound $C(G)$ pick the most useful W with $G(W) = G$.
- That is, define one output symbol for each clique in G . Then

$$R_\infty(W_{\text{opt}}) = \log \bar{\chi}^*(G)$$

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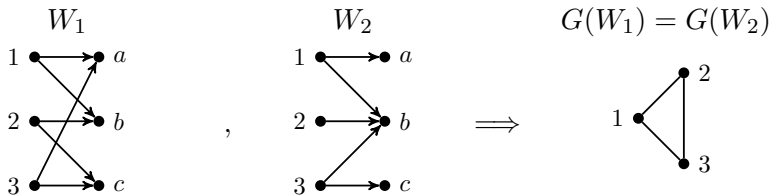


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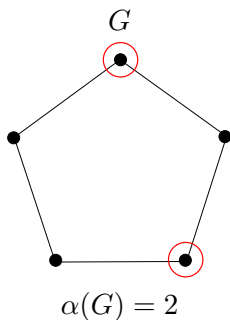
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G^2

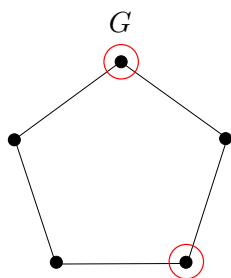
- Achievability:

$$\alpha(G^2) = 5 \implies C(G) \geq \frac{1}{2} \log 5$$

- Converse: fractional clique cover with 5 cliques, weight 1/2 each

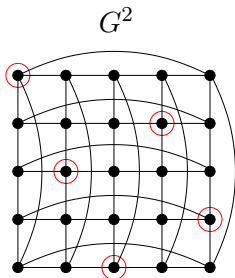
$$\bar{\chi}^*(G) = 5/2 \implies C(G) \leq \log(5/2)$$

- Lovász (1979): $C(G) = \frac{1}{2} \log 5$



$$\alpha(G) = 2$$

but



$$\alpha(G^2) = 5$$

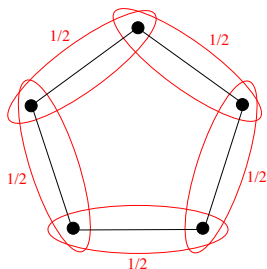
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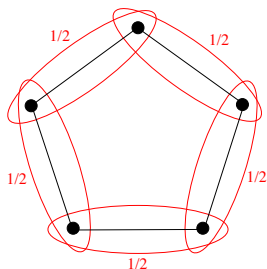
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Lovász's idea

- Graph representation: map vertices x to unit norm $u_x \in \mathbb{R}^d$ so that

$$x \sim x' \implies u_x \perp u_{x'}$$

- An independent set A is mapped to an orthonormal basis
- For any unit norm c and independent set A

$$1 \geq \|c\|^2 \geq \sum_{x \in A} |\langle u_x | c \rangle|^2 \geq |A| \min_x |\langle u_x | c \rangle|^2$$

- Take $\{u_x\}$ and c optimally; if

$$\theta(G) = \left(\max_{\{u_x\}, c} \min_x |\langle u_x | c \rangle|^2 \right)^{-1}$$

then

$$\alpha(G) \leq \theta$$

Tensorization

- Note $\langle a \otimes b | c \otimes d \rangle = \langle a | c \rangle \langle b | d \rangle$
- So, if $\{u_x\}$ representation of G used with c gives

$$\alpha(G) \leq \theta(G)$$

then $\{u_x\}^{\otimes n}$ representation of G^n used with $c^{\otimes n}$ gives

$$\alpha(G^n) \leq \theta(G)^n$$

- Taking $\lim \frac{1}{n} \log(\cdot)$

$$C(G) \leq \log \theta(G)$$

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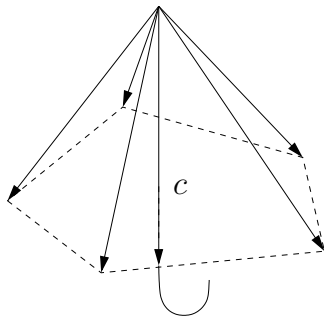
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Lovász's umbrella



$$\theta = \sqrt{5}$$

$$C(G) \leq \log \sqrt{5}$$

- **Orthonormal Representation:**

A set of unit norm vectors $\{u_x\}$, $x \in \mathcal{X}$

$$x, x' \text{ not confusable} \implies \langle u_x | u_{x'} \rangle = 0$$

- **Trivial Representation:** $u_x = \sqrt{W_x}$

- **Value (log domain):**

$$V(\{u_x\}) = \min_c \max_x \log \frac{1}{|\langle u_x | c \rangle|^2} \quad (\|c\| = 1)$$

c is the *handle*. Note: $|\langle u_x | c \rangle|^2 \geq e^{-V(\{u_x\})}$, $\forall x$

- **The bound:**

$$C_0 \leq V(\{u_x\})$$

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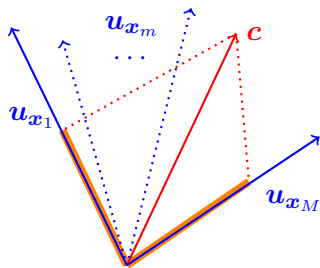
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Representation for \mathbf{W}

Vectors $\mathbf{x} = (x_1, \dots, x_n) \longrightarrow \mathbf{u}_{\mathbf{x}} = u_{x_1} \otimes \dots \otimes u_{x_n}$

Handle $\mathbf{c} = c \otimes \dots \otimes c$

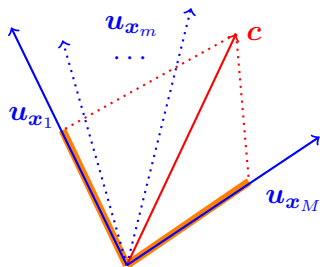


- For a zero-error code, the vectors \mathbf{u}_{x_m} are pairwise orthogonal
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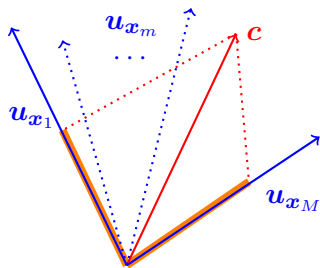


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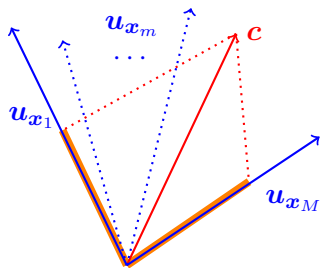


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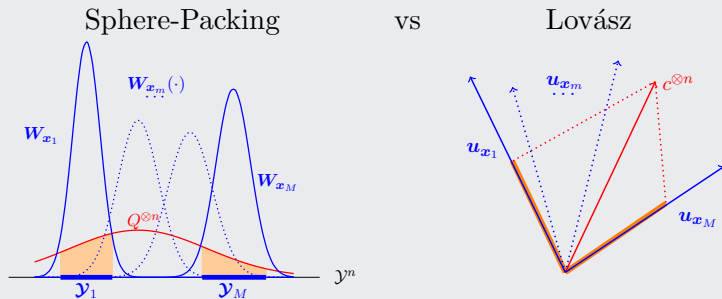
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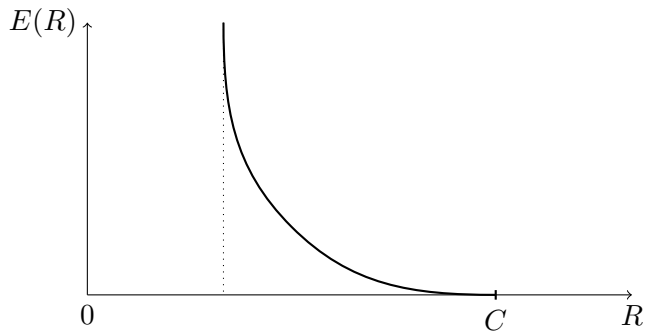
Analogies



We note the following analogies

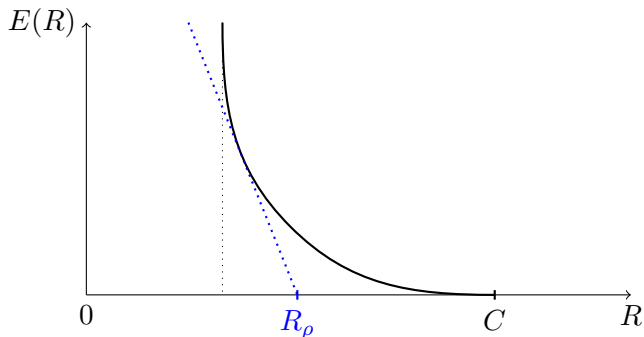
$$\begin{aligned}
 \sqrt{W_{x_m}} &\leftrightarrow \mathbf{u}_{x_m} \\
 \sqrt{Q} &\leftrightarrow \mathbf{c} \\
 Q^{\otimes n}(\mathcal{Y}_m) &\leftrightarrow |\langle \mathbf{u}_{x_m} | \mathbf{c}^{\otimes n} \rangle|^2
 \end{aligned}$$

Sphere-Packing Bound as an Information Radius



What about min-max expressions?

Sphere-Packing Bound as an Information Radius

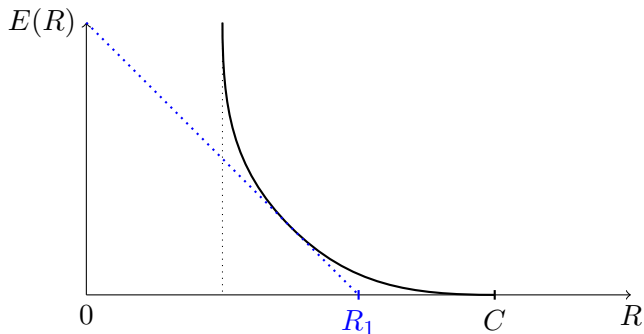


Remind

$$R_\rho = \min_Q \max_x D_\alpha(W_x || Q), \quad \alpha = 1/(1 + \rho)$$

$D_\alpha(Q_1 || Q_2) = \frac{1}{\alpha-1} \log \sum_y Q_1(y)^\alpha Q_2(y)^{1-\alpha}$ is the Rényi divergence

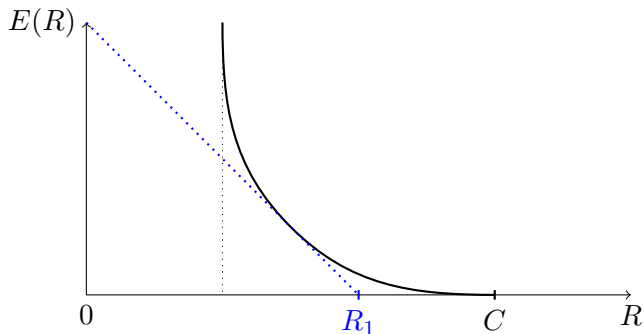
Sphere-Packing Bound as an Information Radius



Setting $\rho = 1$, **cutoff rate**:

$$R_1 = \min_Q \max_x \log \frac{1}{\left(\sum_y \sqrt{W_x(y)Q(y)} \right)^2}$$

Sphere-Packing Bound as an Information Radius



Setting $\rho = 1$, **cutoff rate**:

$$\begin{aligned} R_1 &= \min_Q \max_x \log \frac{1}{\left(\sum_y \sqrt{W_x(y)Q(y)}\right)^2} \\ &= V(\{u_x\}) \quad \text{if } u_x = \sqrt{W(\cdot|x)} \end{aligned}$$

Representations, values and cutoff rates

- So,

$$u_x = \sqrt{W_x} \implies V(\{u_x\}) = \text{cutoff rate}$$

- If all u_x have non-negative components we always get the cutoff rate of some classical channel
- Lovász' optimal u_x can (often will!) have negative components.

Intuition (?)

Use wave functions of quantum physics to play the role of $\sqrt{W_x}$

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Representations, values and cutoff rates

- So,

$$u_x = \sqrt{W_x} \implies V(\{u_x\}) = \text{cutoff rate}$$

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- **Basic Idea**

W_x now density operator

W_x is a positive semi-definite matrix with unit trace

- **Classical channels:** all w_x are diagonal

$$W_x = \begin{bmatrix} W_x(1) & 0 & \cdots & 0 \\ 0 & W_x(2) & \cdots & 0 \\ 0 & \cdots & \ddots & \end{bmatrix}$$

- **Pure-State Channel:** all W_x are rank-one matrices

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- **Memoryless extension:**

$$\mathbf{x} = (x_1, \dots, x_n) \rightarrow \mathbf{W}_{\mathbf{x}} = W_{x_1} \otimes \dots \otimes W_{x_n}$$

- **Code:** M codewords $\{\mathbf{x}_1, \mathbf{x}_2, \dots, \mathbf{x}_M\} \subset \mathcal{X}^n$
- **Decoder:** a POVM, collection of M positive operators $\{\Pi_1, \dots, \Pi_M\}$ (positive semi-definite matrices) such that

$$I - \sum_{m=1}^M \Pi_m \geq 0$$

- Classical deterministic case: Π_m diagonal $\{0, 1\}$ -valued matrix, indicator function of \mathcal{Y}_m
- **Probability of error:** $P_{e|m} = 1 - \text{Tr}(\Pi_m \mathbf{W}_{\mathbf{x}_m})$
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- If $A = |a\rangle\langle a|$ and $B = |b\rangle\langle b|$ (pure states)

$$\text{Tr } AB = |\langle a|b\rangle|^2$$

- If

$$A = \sum_i \alpha_i |a_i\rangle\langle a_i| \quad B = \sum_j \beta_j |b_j\rangle\langle b_j|$$

then

$$\text{Tr } AB = \sum_{i,j} \alpha_i \beta_j |\langle a_i|b_j\rangle|^2$$

Capacity

- Hausladen-Jozsa-Schumacher-Westmoreland-Wootters: (1996) pure states
- Holevo, Schumacher-Westmoreland (1998): general states

$E(R)$ - achievability

- Burnashev-Holevo (1998): random coding for pure states
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- Hayashi (2006): best “random coding” bound for mixed states
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- Consider again binary hypothesis testing
- Try both MIT approach and Harountunian's approach

Quantum Binary Hypothesis Testing

- Here σ_0, σ_1 are density operators, with

$$P_{e|\sigma_0} = \text{Tr} \sigma_0^{\otimes n} (I - \Pi) \quad P_{e|\sigma_1} = \text{Tr} \sigma_1^{\otimes n} \Pi$$

- Error exponents:

$$\begin{aligned} -\frac{1}{n} \log P_{e|\sigma_0} &= -\mu(s) + s\mu'(s) + o(1) \\ -\frac{1}{n} \log P_{e|\sigma_1} &= -\mu(s) - (1-s)\mu'(s) + o(1) \end{aligned}$$

where

$$\begin{aligned} \mu(s) &= \log \text{Tr} \sigma_0^{1-s} \sigma_1^s \\ &= -s D_{1-s}(\sigma_0 \| \sigma_1) \end{aligned}$$

and

$$D_\alpha(\rho \| \sigma) = \frac{1}{\alpha - 1} \text{Tr} \rho^\alpha \sigma^{1-\alpha}$$

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$$D_\alpha(\rho \parallel \sigma) = \frac{1}{\alpha - 1} \text{Tr } \rho^\alpha \sigma^{1-\alpha}$$

- Upon differentiation, one finds for example for $P_{e|\sigma_0}$

$$-\frac{1}{n} \log P_{e|\sigma_0} = -\log \text{Tr}(\sigma_0^{1-s} \sigma_1^s) + \text{Tr} \left[\frac{\sigma_0^{1-s} \sigma_1^s}{\text{Tr} \sigma_0^{1-s} \sigma_1^s} (\log \sigma_1^s - \log \sigma_0^s) \right] + o(1)$$

- When σ_0 and σ_1 commute, define

$$\sigma_s = \frac{\sigma_0^{1-s} \sigma_1^s}{\text{Tr} \sigma_0^{1-s} \sigma_1^s}$$

and use $\log \sigma_1^s - \log \sigma_0^s = \log \sigma_0^{1-s} \sigma_1^s - \log \sigma_0$.

- This gives for example (same for $P_{e|\sigma_1}$)

$$\begin{aligned} -\frac{1}{n} \log P_{e|\sigma_0} &= \text{Tr} \sigma_s (\log \sigma_s - \log \sigma_0) + o(1) \\ &= D(\sigma_s \| \sigma_0) + o(1). \end{aligned}$$

- But if σ_0, σ_1 do not commute, this form does not hold!

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Example

- Non-orthogonal pure states $\sigma_0 = |\psi_0\rangle\langle\psi_0|$ and $\sigma_1 = |\psi_1\rangle\langle\psi_1|$
- Since $\sigma_0^{1-s} = \sigma_0$ and $\sigma_1^s = \sigma_1$

$$\mu(s) = \log |\langle\psi_0|\psi_1\rangle|^2$$

- At least one of the two error exponents is not larger than $-\log |\langle\psi_0|\psi_1\rangle|^2$.
- Thus, error exponents cannot be expressed as $D(\sigma_s\|\sigma_i)$

$$D(\rho\|\sigma_i) = \begin{cases} 0 & \rho = \sigma_i \\ +\infty & \rho \neq \sigma_i \end{cases}, i = 0, 1,$$

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Channel and coding scheme

- W_x are density operators (classical case: diagonal)
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MIT proof

- Extends using quantum Rényi divergence $D_\alpha(\rho\|\sigma)$
- Matches achievability at high rates for pure-state channels
- Auxiliary Q does *not* induce auxiliary channel V

Haroutunian's approach

- Extends using quantum KL divergence
- Trivial bound for pure-state channels:

$$E(R, P) \leq \infty, \quad R < I(P, W)$$

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$$D(\rho||\sigma) = \begin{cases} 0 & \rho = \sigma \\ +\infty & \rho \neq \sigma \end{cases},$$

- If $R < I(P, W)$ then $I(P, V) < I(P, W)$
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- The bound:

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What is the problem here?

What happened

- Using a constant Q we get a good bound
- Using an optimal channel V we don't
- Impossible... a constant Q is a “dummy channel” with $V_x = Q$

MIT Proof

- Dummy Q
- Converse for Q of the form $\text{Tr } Q\Pi_m \leq e^{-nR}$
- Lower bound $\text{Tr } W_{x_m}\Pi_m$ using BHT between Q and W_{x_m} in the regime where both error probabilities vanish exponentially

Haroutunian

- General channel V with $I(P, V) < R$
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Haroutunian

- General channel V with $I(P, V) < R$
- Converse for V of the form $\text{Tr } \mathbf{V}_{x_m} \Pi_m = o(1)$... too weak
- BHT between \mathbf{V}_{x_m} and \mathbf{W}_{x_m} in Stein's regime

What we should do

- Take an auxiliary V with $I(P, V) < R$
- Compute the correct strong converse $\text{Tr } V_{x_m} \Pi_m = e^{-nE_{sc}(R, P)}$
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Classical case

- Choosing $I(P, V) = 0$ (MIT) or $I(P, V) = R - \epsilon$ (Haroutunian) makes no difference
- No other choice can do better (I guess... list decoding)
- The strong converse exponent for V , and the BHT between V_{x_m} and W_{x_m} both involve Rényi divergences

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Classical-Quantum

- Choosing $I(P, V) = 0$ (MIT) or $I(P, V) = R - \epsilon$ (Haroutunian) does make a difference
- Is $I(P, V) = 0$ really optimal?
→ No matching achievability for mixed state channels.
- Strong converse exponent for c-q channels derived only very recently (Mosonyi and Ogawa 2014).
- Unlike the BHT between V_{x_m} and W_{x_m} , strong converse involves so-called “sandwiched” Rényi divergence

$$\tilde{D}_\alpha(\rho, \sigma) = \frac{1}{\alpha - 1} \log \text{Tr} \left(\sigma^{\frac{1-\alpha}{2\alpha}} \rho \sigma^{\frac{1-\alpha}{2\alpha}} \right)^\alpha.$$

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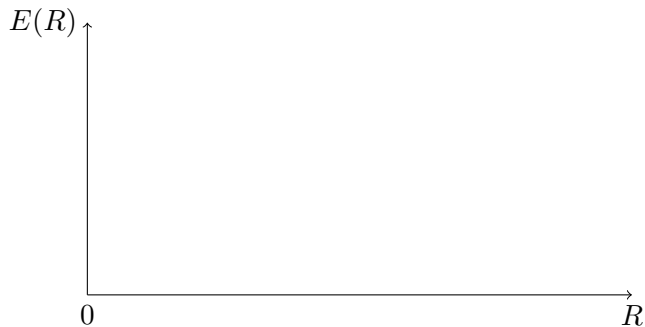
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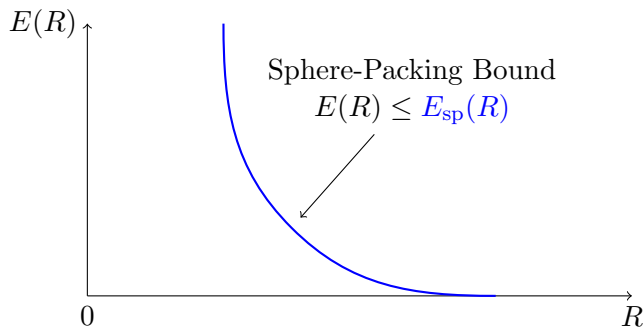
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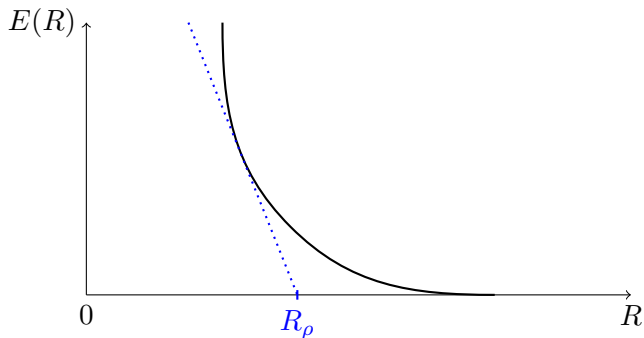
Classical-Quantum Channels: Reliability Function





Sphere packing

$$E_{\text{sp}}(R) = \sup_{\rho \geq 0} \max_P \left[-\log \text{Tr} \left(\sum_x P(x) W_x^{1/(1+\rho)} \right)^{1+\rho} - \rho R \right]$$



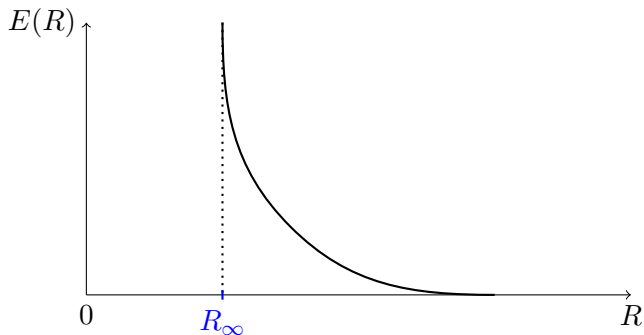
Minmax characterization

$$R_\rho = \min_F \max_x D_\alpha(W_x || F), \quad \alpha = 1/(1 + \rho)$$

where F runs over density operators and

$D_\alpha(F_1 || F_2) = \frac{1}{\alpha-1} \log \text{Tr}(F_1^\alpha F_2^{1-\alpha})$ is the Rényi divergence

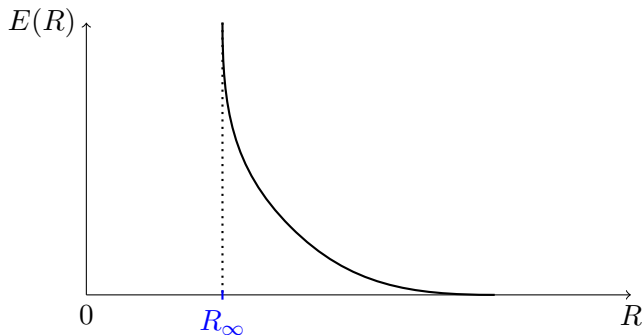
Classical-Quantum Channels: Reliability Function



When $\rho \rightarrow \infty$

$$R_\infty = \min_F \max_x \log \frac{1}{\text{Tr}(W_x^0 F)}$$

where W_x^0 is the projector onto the support of W_x



For pure-state channels $W_x = |u_x\rangle\langle u_x|$

Using pure-states $F = |f\rangle\langle f|$ we have $\text{Tr}(W_x^0 F) = |\langle u_x | f \rangle|^2$.

So,

$$\begin{aligned} R_\infty &\leq \min_f \max_x \log \frac{1}{|\langle u_x | f \rangle|^2} \\ &= V(\{u_x\}) \end{aligned}$$

Orthonormal Representations and Auxiliary Channels

- For any representation $\{u_x\}$, the classical-quantum channel with pure-states $W_x = |u_x\rangle\langle u_x|$ satisfies $R_\infty \leq V(\{u_x\})$

- We can define

$$\vartheta_{sp} = \min_{\{W_x\}} R_\infty \quad (1)$$

where we minimize over all channels such that $\text{Tr } W_x W_{x'} = 0$ if x and x' are not confusable

- Then

$$C_0 \leq \vartheta_{sp} \leq \vartheta$$

- Actually additional results in Lovász' paper imply $\vartheta \leq \vartheta_{sp}$ and hence $\vartheta_{sp} = \vartheta$.
- So, pure-state channels achieve the optimum in (1) and for some optimal channel some pure state $F = |f\rangle\langle f|$ achieves R_∞

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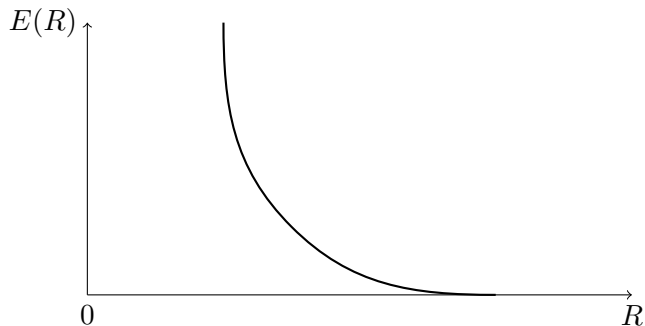
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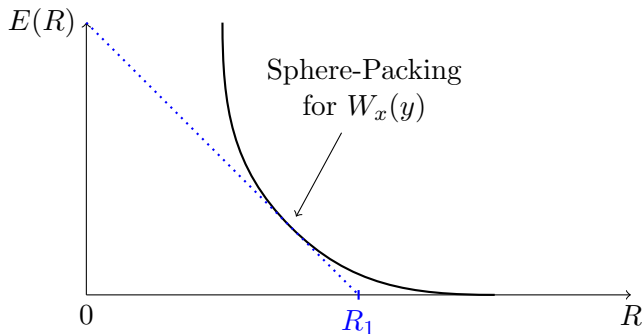
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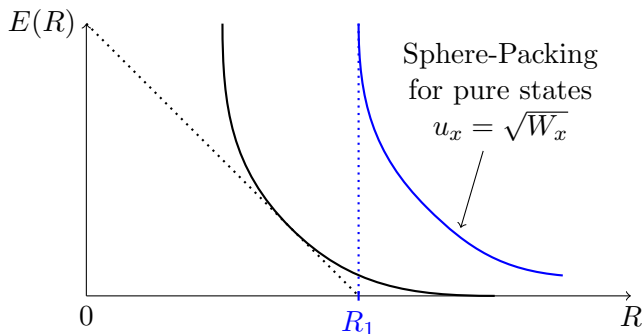


But... where are those cutoff rates?



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- We had previously identified R_1 with $V(\{\sqrt{W_x}\})$
- But then we ended up with a relation between ϑ and R_∞








But... where are those cutoff rates?






- Mathematically, this is due to the fact that the cutoff rate of a channel W always equals the R_∞ rate of a pure-state channel with state vectors $u_x = \sqrt{W_x}$
- The true meaning of this... I do not know, but this sounds important

Suggested reading (to start with) I

-  C. E. Shannon. “The Zero-Error Capacity of a Noisy Channel”. In: *IRE Trans. Inform. Theory* IT-2 (1956), pp. 8–19.
-  C. E. Shannon. “Certain results in coding theory for noisy channels”. In: *Information and Control* 1 (1957), pp. 6–25.
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-  C. E. Shannon, R. G. Gallager, and E. R. Berlekamp. “Lower Bounds to Error Probability for Coding in Discrete Memoryless Channels. I”. In: *Information and Control* 10 (1967), pp. 65–103.

Suggested reading (to start with) II

-  C. E. Shannon, R. G. Gallager, and E. R. Berlekamp. “Lower Bounds to Error Probability for Coding in Discrete Memoryless Channels. II”. In: *Information and Control* 10 (1967), pp. 522–552.
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-  A. J. Viterbi and J. K. Omura. *Principles of Digital Communication and Coding*. McGraw-Hill, New York, 1979.
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M. Nussbaum and A. Szkoła. “The Chernoff lower bound for symmetric quantum hypothesis testing”. In: *Ann. Statist.* 37.2 (2009), pp. 1040–1057.



H. Nagaoka. “The Converse Part of the Theorem for Quantum Hoeffding Bound”. In: *arXiv:quant-ph/0611289v1* ().