

Introduction

• Goal: looking for a practical coding method for a two-user Gaussian broadcast channel

• **Broadcast channel:** simultaneous communication of a single source with multiple receivers

• Bergsman theorem: for some p(x, v), capacity region of a degraded BC is

$$R_z \leq I(V;Z), \quad R_y \leq I(X;Y|V)$$

• Gaussian broadcast channel

$$Y = AX + N_y, \quad Z = BX + N_z$$

- a power constraint input $\mathbb{E}(|X|^2) \leq P$
- wlog, for fixed $|A| > |B| \implies$ degraded BC

$$\bigcup_{\alpha \in [0,1]} \left\{ \begin{array}{l} R_y \leq \frac{1}{2} \log_2 \left[1 + \alpha |A|^2 \frac{P}{N} \right] \\ R_z \leq \frac{1}{2} \log_2 \left[1 + \frac{(1-\alpha)|B|^2 P}{N+\alpha|B|^2 P} \right] \end{array} \right\}$$

• α is the fraction of input power allocated to user *Y*



• boundary is achieved by a Gaussian codebook

 $X = \sqrt{\alpha P} X_{y} + \sqrt{\bar{\alpha} P} X_{z} \quad (X_{y}, X_{z}) \sim \mathcal{N}(\mathbf{0}, \mathbf{I}_{2})$

Bad news: cannot have a Gaussian codebook in practice!

Disjoint LDPC Coding for Gaussian Broadcast Channels

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Using Binary Codes

- **good news:** using a binary codebook at low SNRs, we will not lose too much! [1]
- (X_y, X_z) is uniformly picked from $\{+1, -1\}^2$

$$X = \sqrt{\alpha P} X_y + \sqrt{\bar{\alpha} P} X_z$$

• each user has an LDPC ensemble



- optimal MAP detection at user *Y*
- mapper node handles $X = \sqrt{\alpha P} X_{y} + \sqrt{\bar{\alpha} P} X_{z}$
- mapper node is in essence an *interference canceler* [1]

Drawbacks

- joint decoding at both users
- both codes are required at the receivers

Question What if we could get rid of the mapper node?



Bit-Interleaved Coded Modulation (BICM)

is used.

Proposed Method

- and

• Caire [2] BICM performs extremely close to the optimal decoder with a lower complexity if Gray labeling

• using Gray labeling, there is a minor dependency among label bits

• superposition coding is a 4-PAM-like modulation with binary labeling

• LDPC codes are self-interleaved • can apply Gray labeling

$$X = \sqrt{\alpha P} X_{y} + \sqrt{\overline{\alpha} P} X_{z} X_{y}, \ \alpha \ge \frac{1}{2} \quad X_{z} \quad X_{y}$$

$$01 \quad 11 \quad 10 \quad 00$$

$$-\sqrt{\alpha} - \sqrt{\overline{\alpha}} \quad \sqrt{\overline{\alpha}} - \sqrt{\alpha} \sqrt{\alpha} - \sqrt{\overline{\alpha}} \quad \sqrt{\alpha} + \sqrt{\overline{\alpha}}$$

$$X = \sqrt{\alpha P} X_{y} X_{z} + \sqrt{\overline{\alpha} P} X_{z}, \ \alpha \le \frac{1}{2} \quad X_{y} \quad X_{z}$$

$$01 \quad 11 \quad 10 \quad 00$$

$$-\sqrt{\alpha} - \sqrt{\overline{\alpha}} \quad \sqrt{\alpha} - \sqrt{\overline{\alpha}} \quad \sqrt{\alpha} - \sqrt{\overline{\alpha}} \quad \sqrt{\alpha} + \sqrt{\overline{\alpha}}$$

• trying to reduce the dependency of code bits • does not exactly match superposition coding • may get a non-convex region due to nonlinearity • due to minor dependency, we propose to remove mapper nodes



• each user only needs to have its own code • no need for joint decoding • compared to the optimal method, decoding complex-

ity is decreased at least by a factor of 50%



• most of the region is covered by the proposed method • as expected, the region is not convex since we adaptively force labeling to be Gray!

Simulation Results



References



[1] P. Berlin and D. Tuninetti, "LDPC codes for fading gaussian broadcast channels," IEEE Trans. Inf. The*ory*, vol. 51, no. 6, pp. 2173–2182, Jun. 2005. [2] G. Caire, G. Taricco, and E. Biglieri, "Bit-interleaved

coded modulation," IEEE Trans. Inf. Theory, vol. 44, no. 3, pp. 927–946, May 1998.